

Oct. 2024

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# 可重構智慧表面(RIS)OTA量測

-邱宗文 博士 (川升股份有限公司 創辦人)

# 講者介紹



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台北工專電機科  
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## 經歷

- 第四屆臺灣天線工程師學會理事長
- 成功/逢甲/台北大學等兼任助理教授
- 資深科大/技術學院評鑑委員
- 連展無線通訊事業處處長
- 川升股份有限公司創辦人&現任總經理
- 至高頻科技/正于微波共同創辦人
- 2022低軌衛星工業局科專計畫負責人
- 2022國家發明獎/經濟部創新研究獎獲獎人
- 約200項發明專利發明人
- 川升學苑創辦人



## 專長

- 天線OTA量測、天線設計及微波電路應用

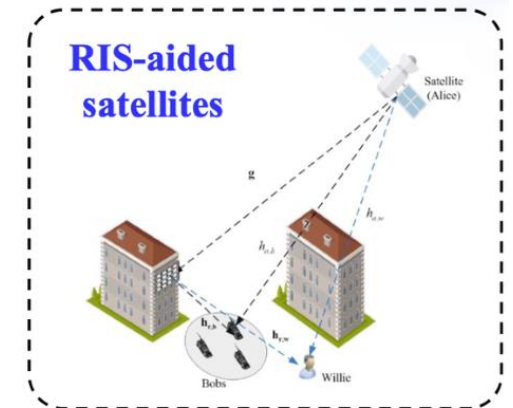
# 大綱

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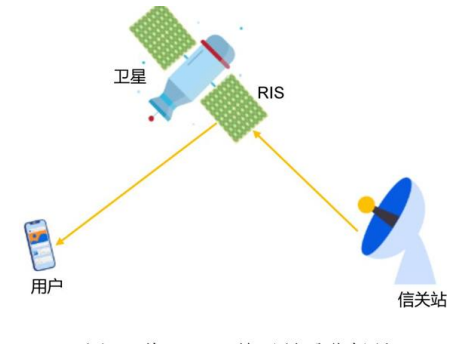
- **RIS及OTA量測重要參數介紹**
- RIS OTA量測文獻分享
- RIS OTA量測系統介紹
- 結論

# Why RIS?

- 5G sub6/mmWave
  - Blockages
  - Severe path loss
  - Channel capacity
- Satellite/UAV
  - Transponder
  - Enhance Aperture size
  - Stealth
- RIS (Reconfigurable Intelligent Surfaces)
  - Cost-effective
  - Power-efficient



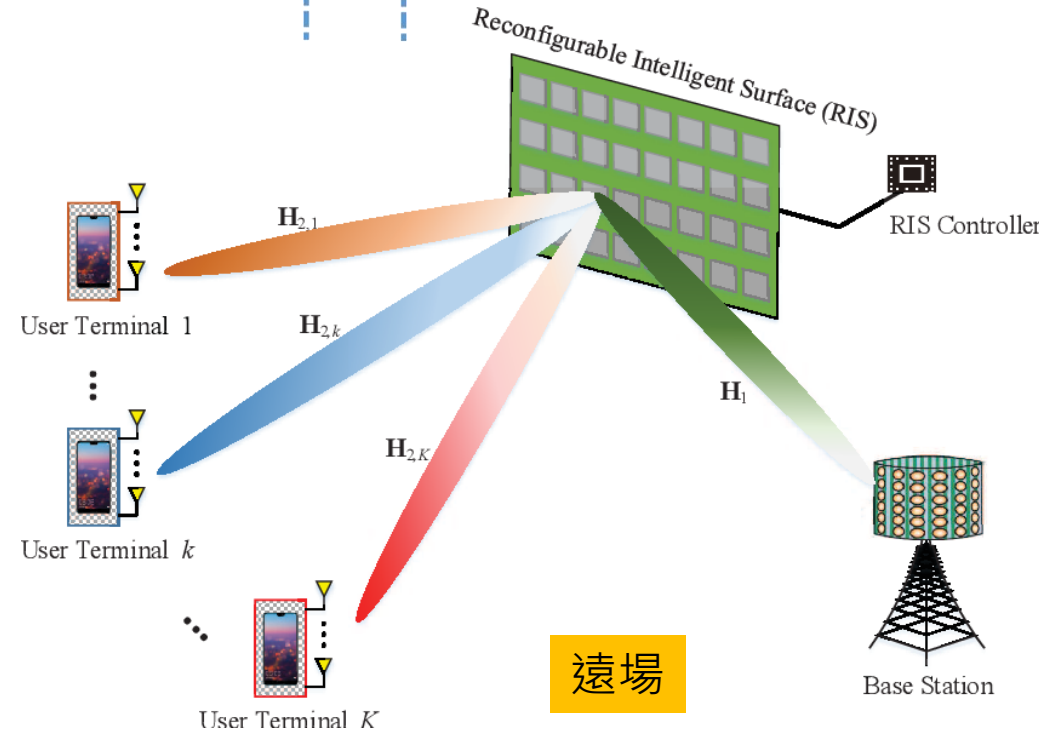
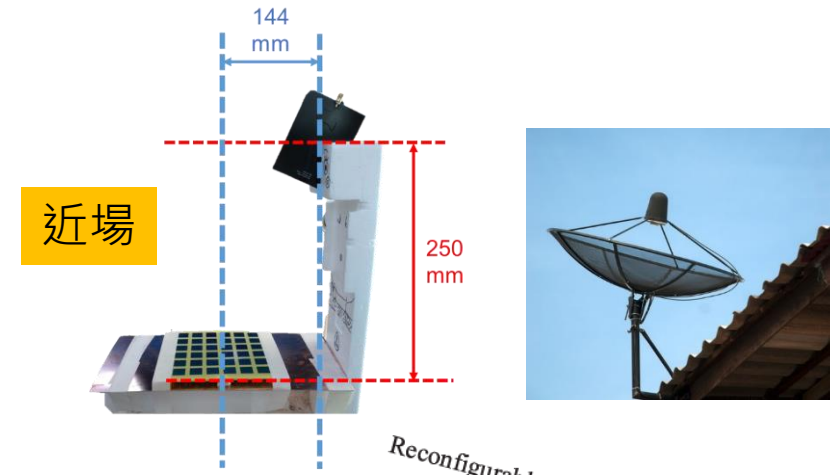
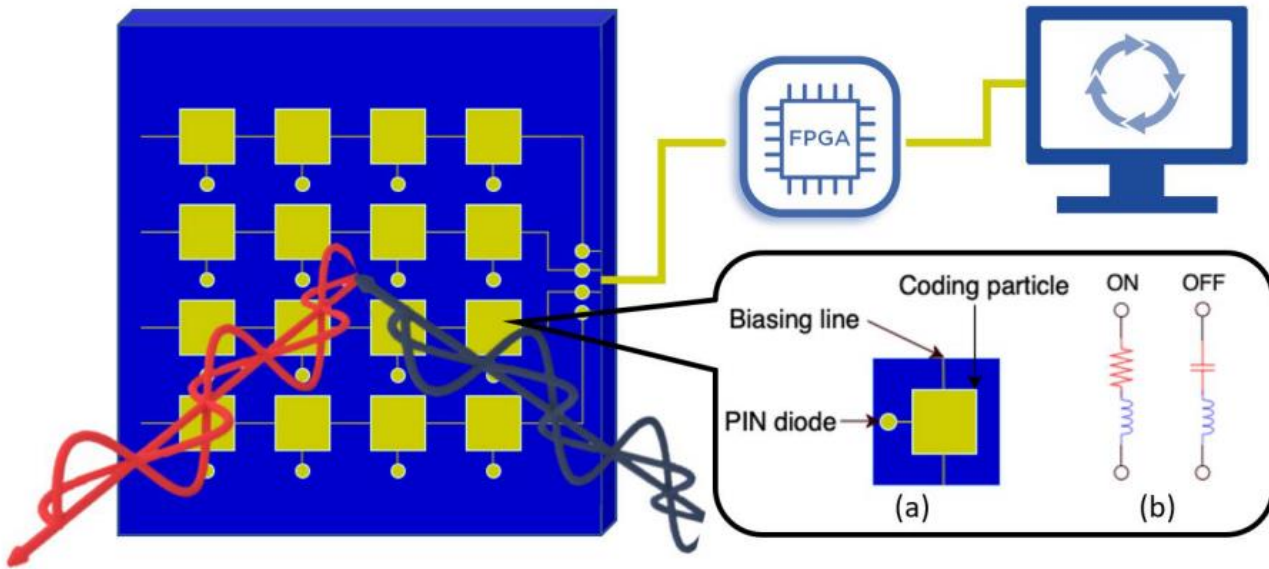
增強收訊/匿蹤



資料來源：網路

# What is RIS?

- Planar surface structure (NF & FF)
- Changing the directions of reflected/transmitted signal
- Reconfiguring the wireless propagation environment



L. You *et al.*, "Reconfigurable Intelligent Surfaces-Assisted Multiuser MIMO Uplink Transmission With Partial CSI," in *IEEE Transactions on Wireless Communications*, vol. 20, no. 9, pp. 5613-5627, Sept. 2021

C. Pan *et al.*, "Reconfigurable Intelligent Surfaces for 6G Systems: Principles, Applications, and Research Directions," in *IEEE Communications Magazine*, vol. 59, no. 6, pp. 14-20, June 2021

[https://xsj.699pic.com/pay/index?click\\_type=851](https://xsj.699pic.com/pay/index?click_type=851)

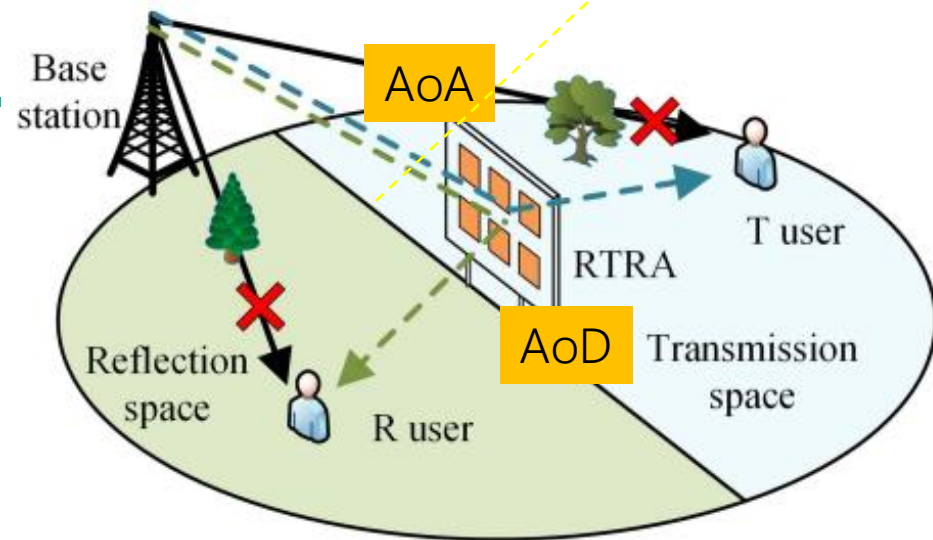
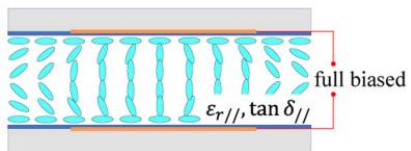
# RIS設計原理



Phase Shift

0/180度  
介電常數

智慧化!



反射:  $\text{Phase} + \text{Phase} + \text{Phase Shift}$

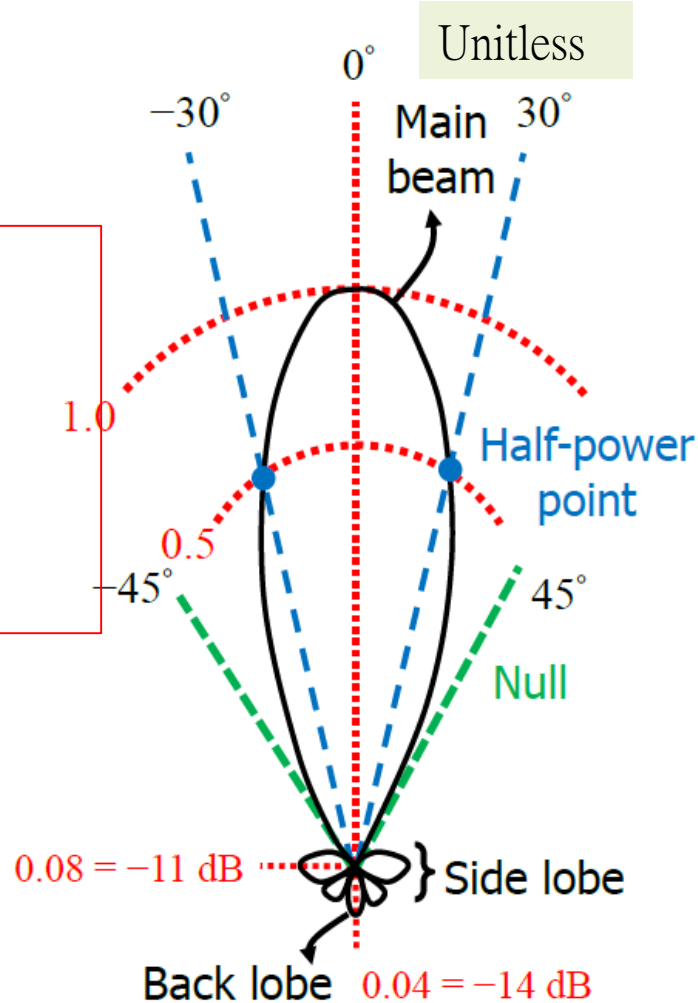
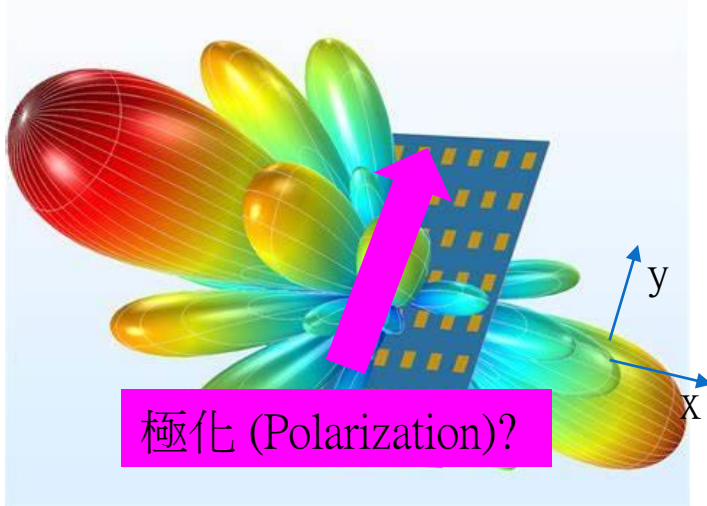
穿透:  $\text{Phase} + \text{Phase} + \text{Phase Shift}$

相等!

# 重要參數-輻射場型

E-plane/H-plane  
Co. pol./X-pol

Directivity : peak/aver.  
HPBW (Half power beamwidth)  
SLL (side lobe level)  
F/B (front to back ratio)  
XPL (cross pol. Level)



Monostatic/Bistatic!

Terminologies concerned about the pattern:

Antenna boresight	The direction of maximum radiation
Half-power beam width (HPBW)	The angle between the half-power point (AKA 3 dB beam width)
Beam width between first nulls (FNBW)	The angle between first nulls for main beam
Front-to-back ratio (F/B ratio)	The ratio of power gain between the front and rear of a directional antenna
Side lobe level (SLL)	The power level of the peak of the side lobe (expressed in decibels)

OTA SNR!!!

# 重要參數 - Aperture Efficiency

Gain

$$A_{\text{eff}} = \frac{G \lambda^2}{4\pi}$$

Wavelength

Effective Aperture

$$\epsilon_a = \frac{A_{\text{eff}}}{A_{\text{phys}}}$$

天線尺寸

Aperture Efficiency

- 2.4 GHz
- 24 inch diameter (0.2235 square meters area)
- 21 dBi

$$\epsilon_a = \frac{10^{21\text{dBi}} \cdot 10 \cdot c^2}{4\pi \cdot 0.2235 \text{ m}^2 \cdot (2.4 \text{ GHz})^2} = 70\%$$

IEEE TRANSACTIONS ON ANTENNAS AND PROPAGATION, VOL. 63, NO. 11, NOVEMBER 2015

## Design of a Wideband Dual-Polarization Full-Corporate Waveguide Feed Antenna Array

Shi-Gang Zhou, Guan-Long Huang, Tan-Huat Chio, Jiang-Jun Yang, and Gao Wei

Moreover, the aperture efficiency is estimated by

$$\eta_{AE} = \frac{G \lambda_0^2}{4\pi A} \times 100\% \quad (1)$$

Feed Networks	Volume Size ( $\lambda_c^3$ )	Central Frequency (GHz)	Impedance & AR Bandwidth (GHz)	Peak Gain (dBic)	Aperture Efficiency
Conventional	2.9×1.8×0.6	8.25	8.15-8.35	16.1	62.5%
		14.5	14.2-14.8	15.2	16.4%
Conventional, PRGW lines	2.3×1.7×0.3	20	19.4-20.4	13.2	59.4%
		30	29.5-30.9	11.5	18.8%
CP horn	14.2×14.2×0.6	12.5	12.2-14.1*	25.3	13.8%
		30	27.3-31.6*	31	15.6%
Sequential rotation	2.5×2.5×0.08	7.55	7.4-7.7	14	29.6%
		8.15	7.95-8.32	14.2	29.2%
Sequential rotation, Waveguide	5.6×5.6×4.7	20	18-22	23.94	61.4%
		30	28-32	26.87	53.6%

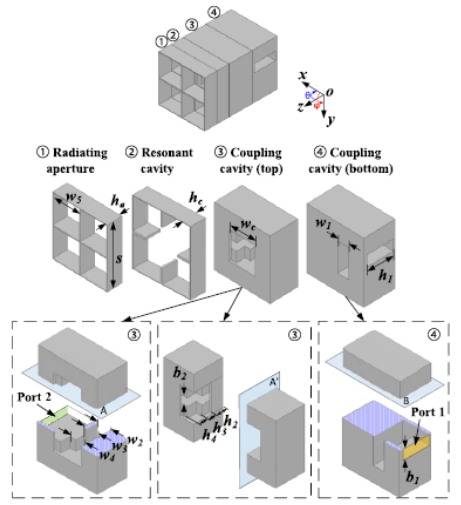


Fig. 2. Configuration of the 2 × 2-element subarray.

lower operating band of the corresponding antenna. ana. which has no impedance bandwidth due to additional horn antenna as its feed.

AE of RIS : <10%



# 重要參數 - RCS

影響雷達探測距離其基本的公式

$$P_r = \frac{P_t G_t}{4\pi r^2} \sigma \frac{1}{4\pi r^2} A_{eff}$$

其中

- $P_t$  = 雷達的發射功率 (單位：瓦特W)
- $G_t$  = 雷達天線增益 (單位：分貝db)
- $r$  = 雷達到探測目標的距離 (單位：公尺M)
- $\sigma$  = 目標的雷達截面積 (單位：RCS平方米)
- $A_{eff}$  = 接收天線的有效面積 (單位：平方米)
- $P_r$  = 接收到的雷達功率 (單位：瓦特W)

#	目標	RCS雷達截面積 (m <sup>2</sup> )
1	飛機	
1.1	戰鬥機	3-12 <sup>[1]</sup>
1.2	經匿蹤處理的戰鬥機	0.3-0.4 <sup>[1]</sup>
1.3	戰術轟炸機	7-10
1.4	重型轟炸機	13-20
1.4.1	戰略轟炸機 B-52	100 <sup>[2]</sup>
1.4	運輸機	40-70
2	艦艇	
2.1	水下航行的潛艦	0 <sup>[3]</sup>
2.2	獨木舟	50
2.3	飛彈快艇	500
2.4	驅逐艦	10000
2.5	航空母艦	50000 <sup>[4]</sup>
3	地面目標	
3.1	汽車	3-10(波長1公分) <sup>[5]</sup>
3.2	T-90主力戰車 (波長 3-8mm)	29 <sup>[6][7]</sup>

# 大綱

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- RIS及OTA量測重要參數介紹
- **RIS OTA量測文獻分享**
- RIS OTA量測系統介紹
- 結論

# 參考文獻 – Open Site

IEEE TRANSACTIONS ON ANTENNAS AND PROPAGATION, VOL. 72, NO. 3, MARCH 2024

235

## A Wideband Reconfigurable Intelligent Surface for 5G Millimeter-Wave Applications

Ruiqi Wang<sup>1b</sup>, Yiming Yang<sup>1b</sup>, *Graduate Student Member, IEEE*, Behrooz Makki<sup>1b</sup>, *Senior Member, IEEE*, and Atif Shamim<sup>1b</sup>, *Fellow, IEEE*

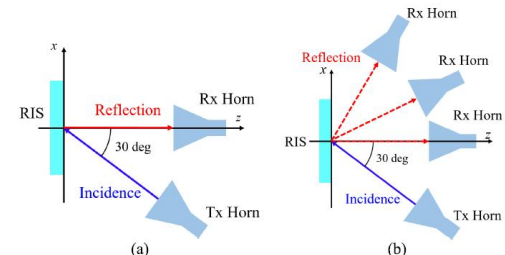
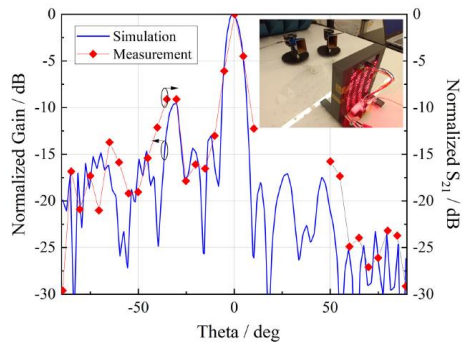
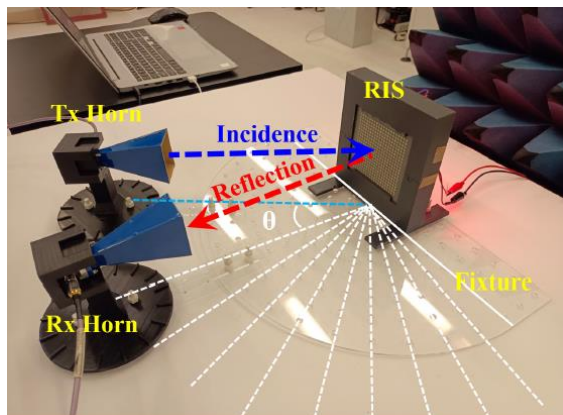
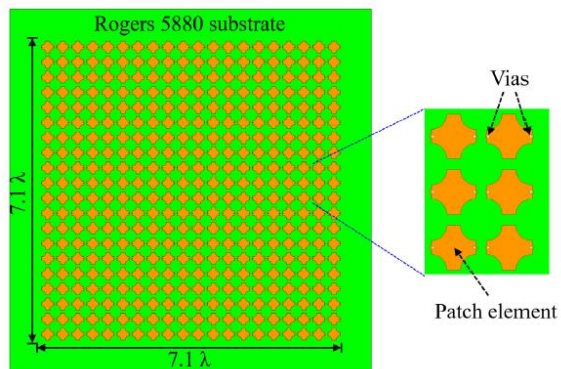
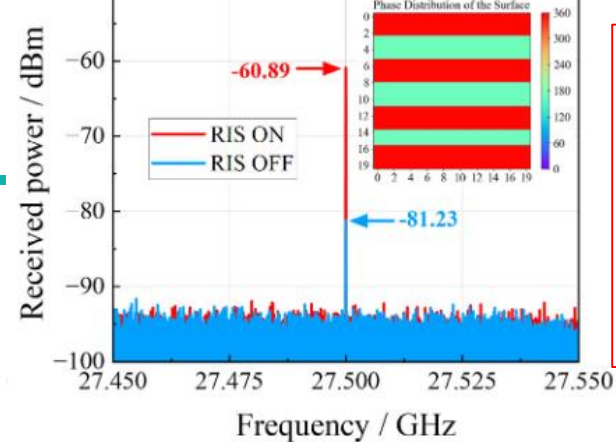
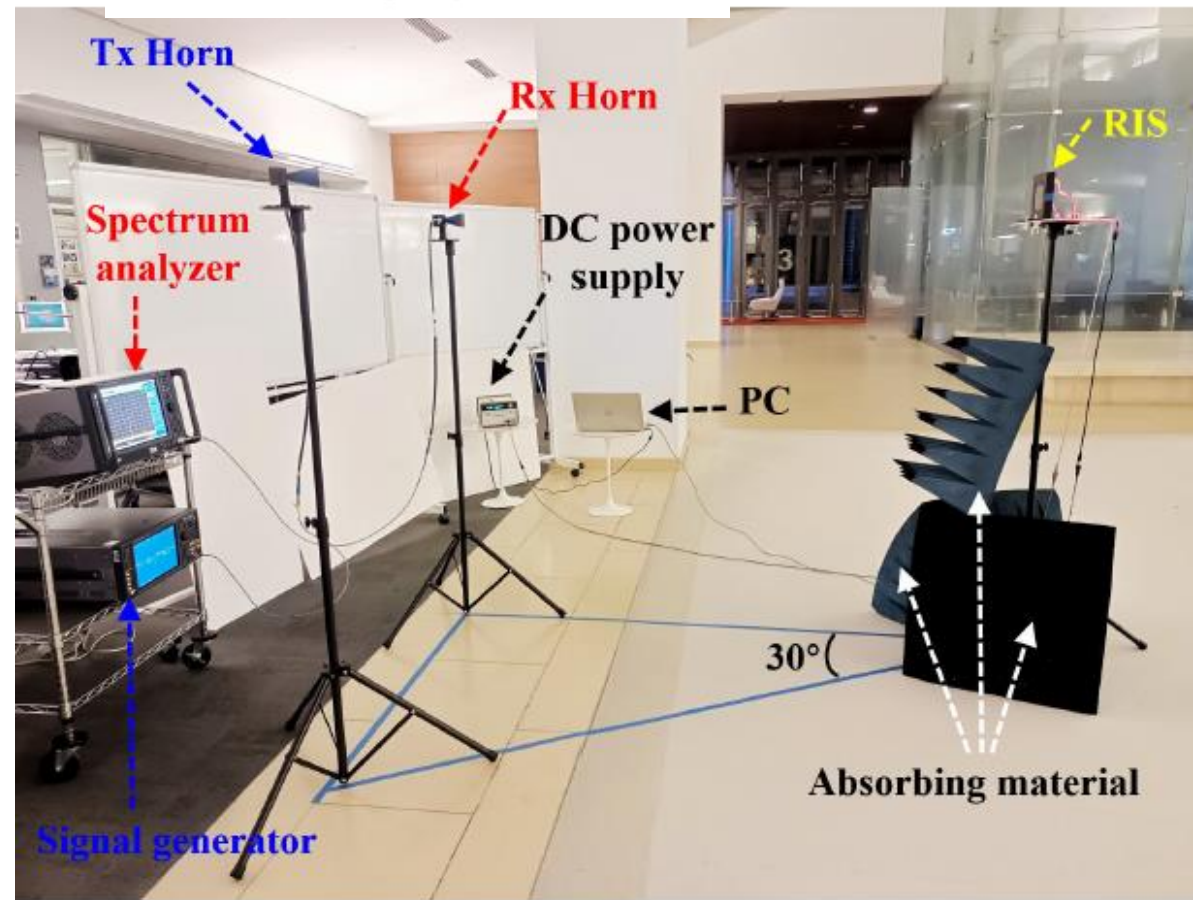


Fig. 17. Measurement scheme. (a) Wideband performance of the proposed RIS. (b) Beam scanning performance of the proposed RIS.



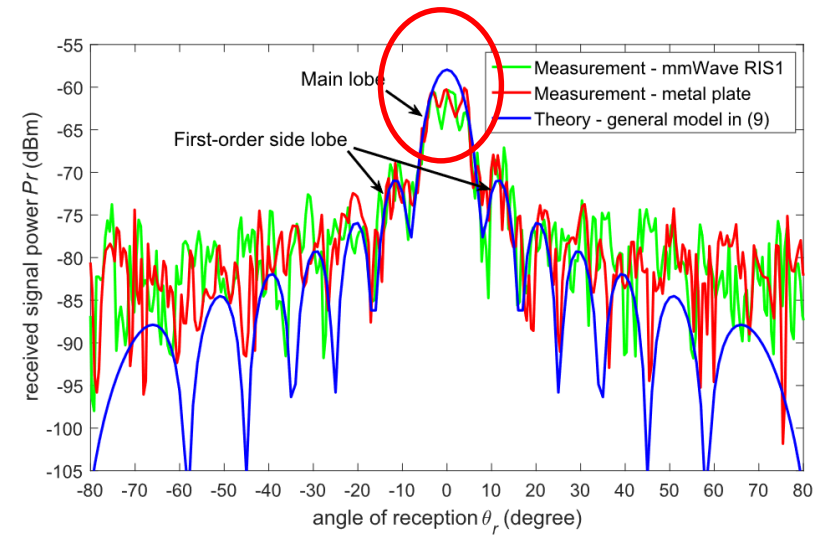
- 目前最常見方法
- 環境干擾
- 單角度量測
- QZ
- Tx/Rx Isolation?



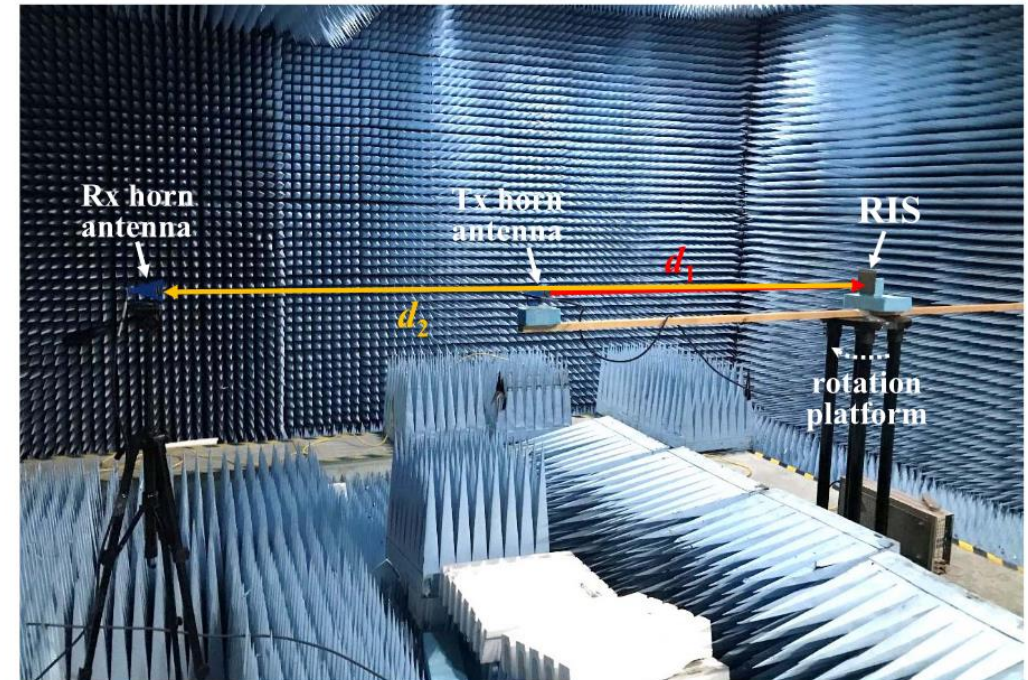
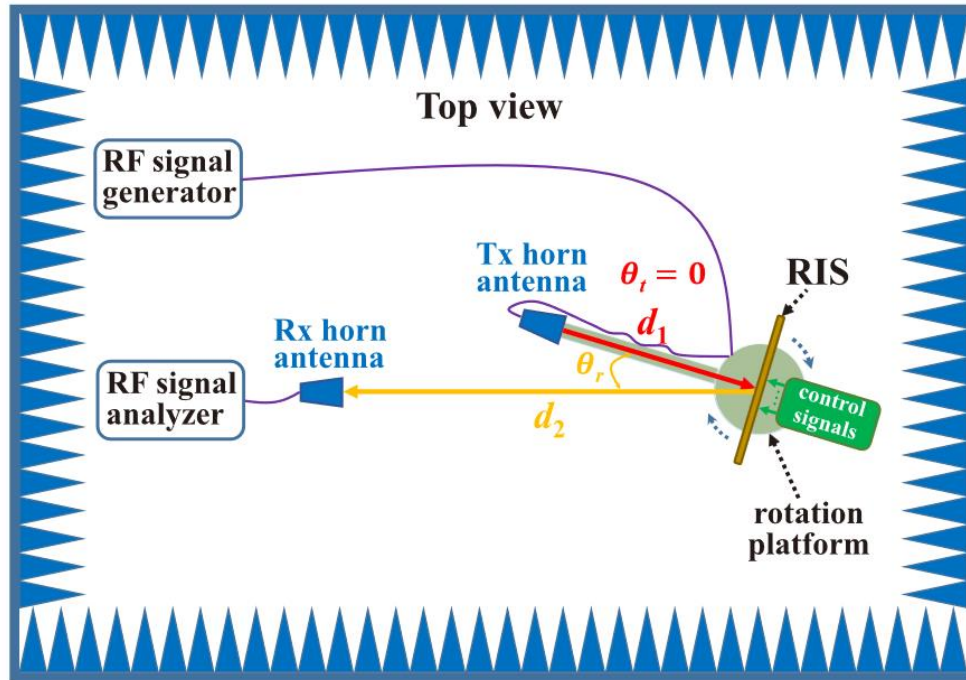
# 參考文獻 – Anechoic Chamber

IEEE TRANSACTIONS ON COMMUNICATIONS, VOL. 70, NO. 9, SEPTEMBER 2022

## Path Loss Modeling and Measurements for Reconfigurable Intelligent Surfaces in the Millimeter-Wave Frequency Band



- 直接遠場 \* 2
- AoA固定
- 2D輻射場型
- AoA/AoD相同?
- Tx/Rx Isolation?



# 參考文獻 – Path Loss校正

IEEE TRANSACTIONS ON COMMUNICATIONS, VOL. 70, NO. 9, SEPTEMBER 2022

System Name	Operating Frequency	Measured Calibration Parameter
Measurement system A	$f = 27$ GHz	$G_t G_r G_{line} = 2.9$ dB
Measurement system B	$f = 27$ GHz	$G_t G_r G_{line} = 24.0$ dB
Measurement system B	$f = 33$ GHz	$G_t G_r G_{line} = 22.0$ dB

- $G_{line}$
- 相位中心
- 極化
- 相對位置
- Multi-path
- QZ

To compare the measurements with the proposed path loss model, the end-to-end antenna gain in (9), i.e.,  $G_t G_r$ , is replaced with the corresponding value obtained from the calibration measurement, i.e.,  $G_t G_r G_{line}$

$$P_r = P_t \frac{G_t G_r \lambda^2}{(4\pi d)^2} G_{line}$$

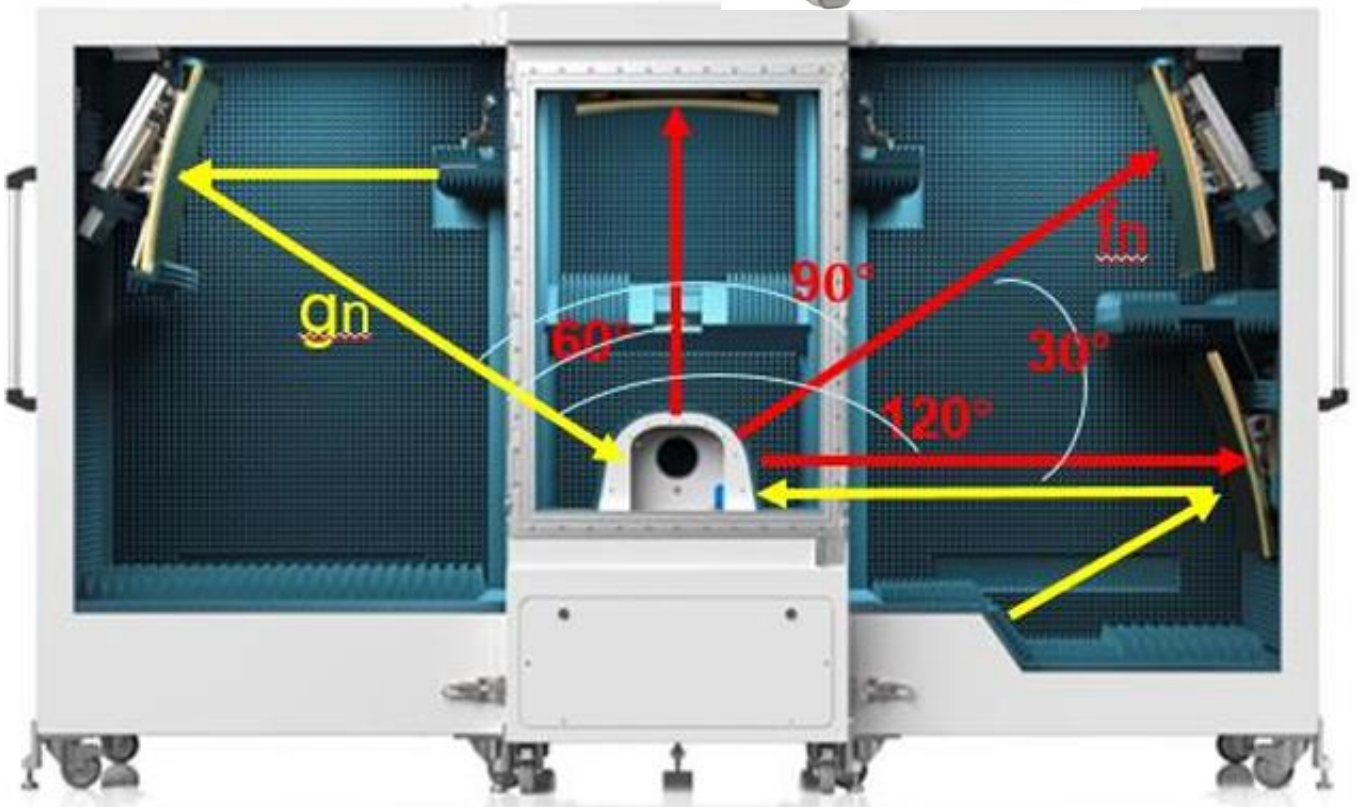
$$PL_{general}^{refined} = \frac{P_t}{P_r} = \frac{16\pi^2}{\underline{G_t G_r} (d_x d_y)^2 \left| \sum_{m=1}^M \sum_{n=1}^N \frac{\sqrt{F_{n,m}^{combine}} \Gamma_{n,m}}{r_{n,m}^t r_{n,m}^r} e^{-j2\pi(r_{n,m}^t + r_{n,m}^r) / \lambda} \right|^2} \quad (9)$$

# 參考文獻 - Active

- 雙CATR
- 特定AoA/AoD
- EIRP/EVM/ACLR

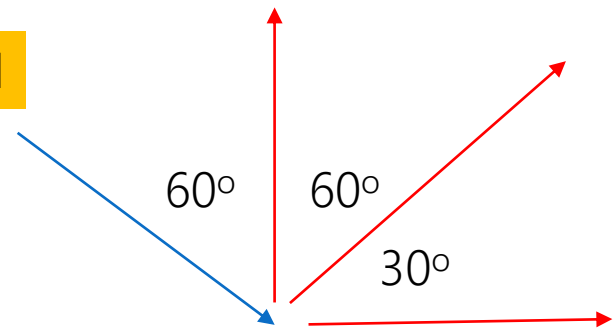
SMW200A

FSW

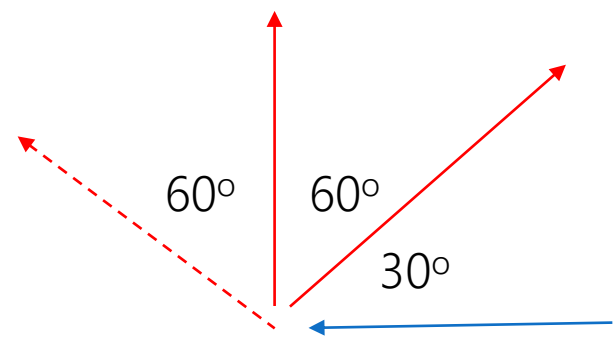


60度/120度/150度

入射1



30度/90度



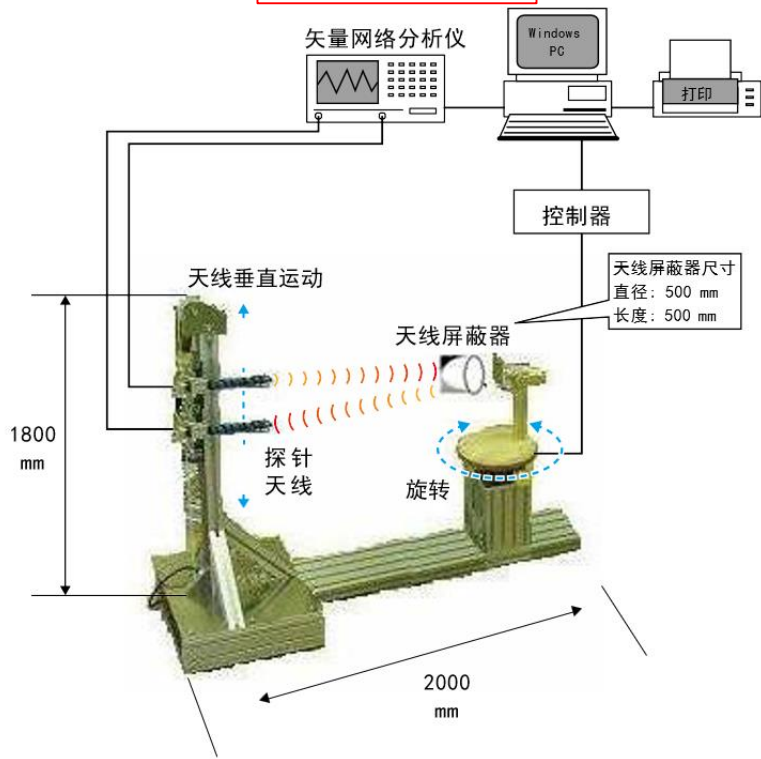
入射2

資料來源：2023 RISTA技術白皮書

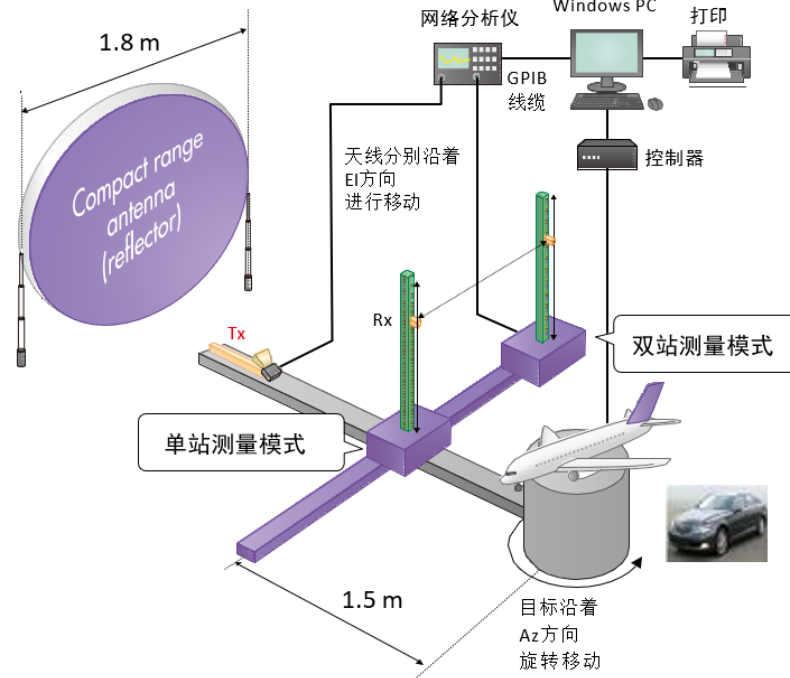
# 參考文獻 – RCS量測

- AoA固定
- 2D輻射場型
- AoA/AoD相同?
- Tx/Rx Isolation?

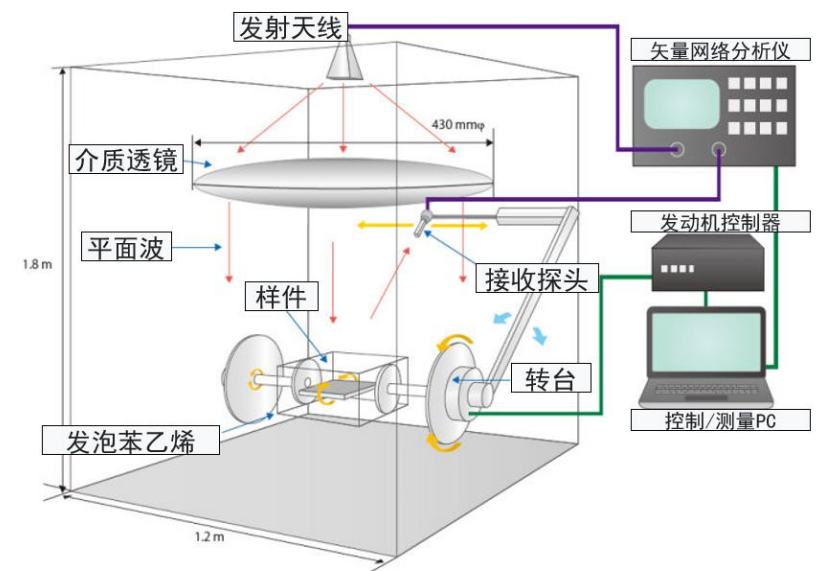
直接遠場\*2



CATR + 直接遠場



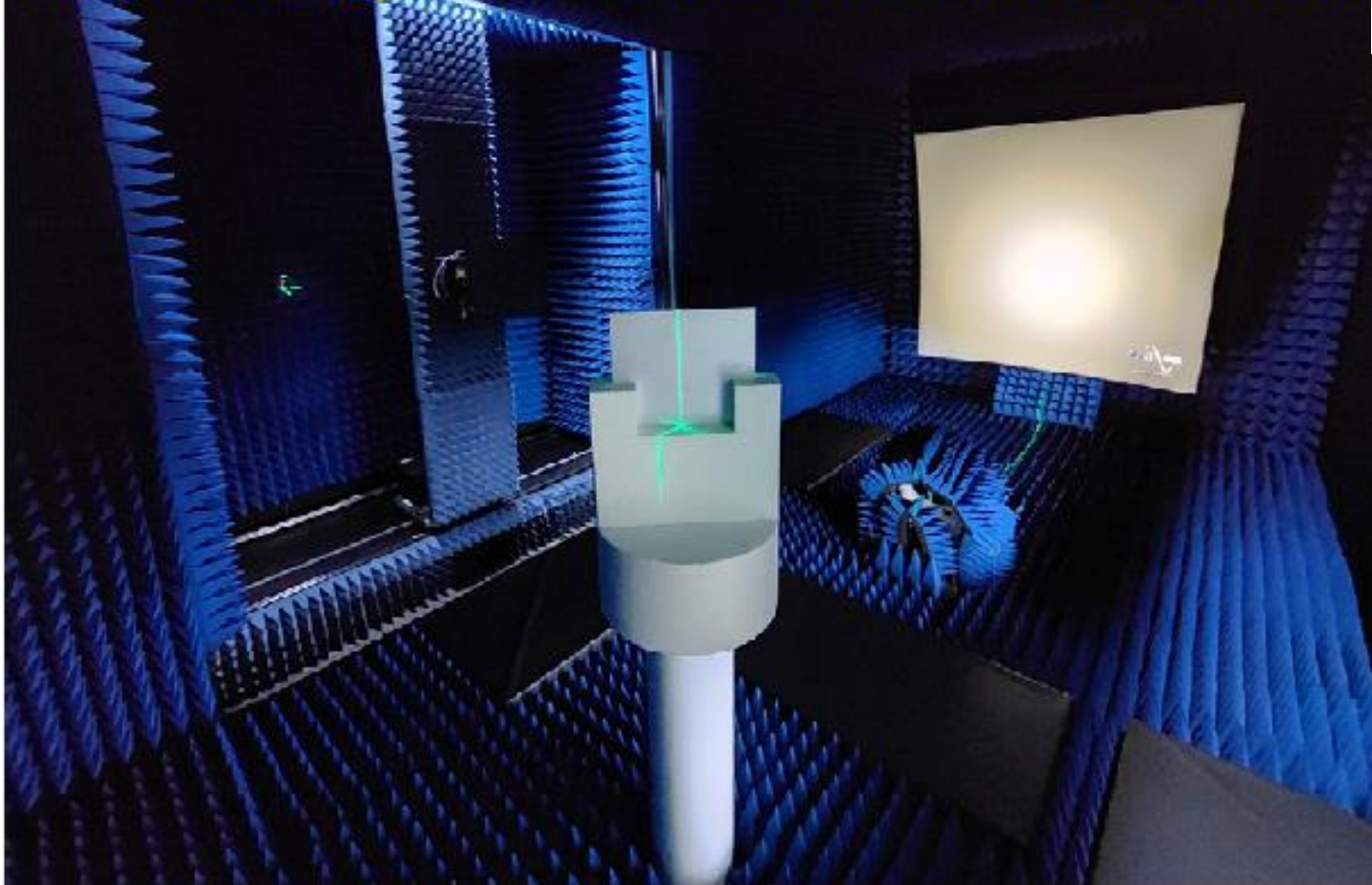
CATR(Lens) + 直接遠場



<https://keycom.co.jp/cmap/xrc.html>

# 參考文獻 – CATR + NF

## Configuration/Scenarios for Bistatic Scattering Properties Evaluation through Planar Near-field to Far-field Transformation



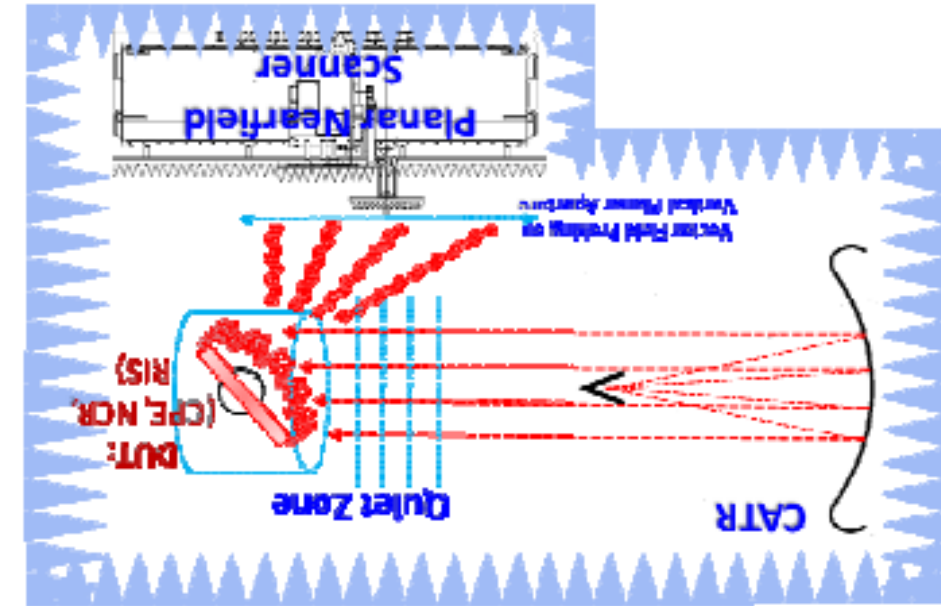
## Performance Evaluation of RU and RIS Based on OTA Mode Near Field and Bistatic Measurement Systems

Chang-Lun Liao <sup>1,2</sup>, Student Member, AMTA, You-Hua Lin <sup>1,3</sup>, Student Member, AMTA, Ike Lin <sup>3</sup>, Member, AMTA, and Chang-Fa Yang <sup>1</sup>, Member, AMTA.

<sup>1</sup> Department of Electrical Engineering, National Taiwan University of Science and Technology (Taiwan Tech), Taiwan

<sup>2</sup> Wireless Communications Laboratory, Chunghwa Telecom Laboratories, Taiwan

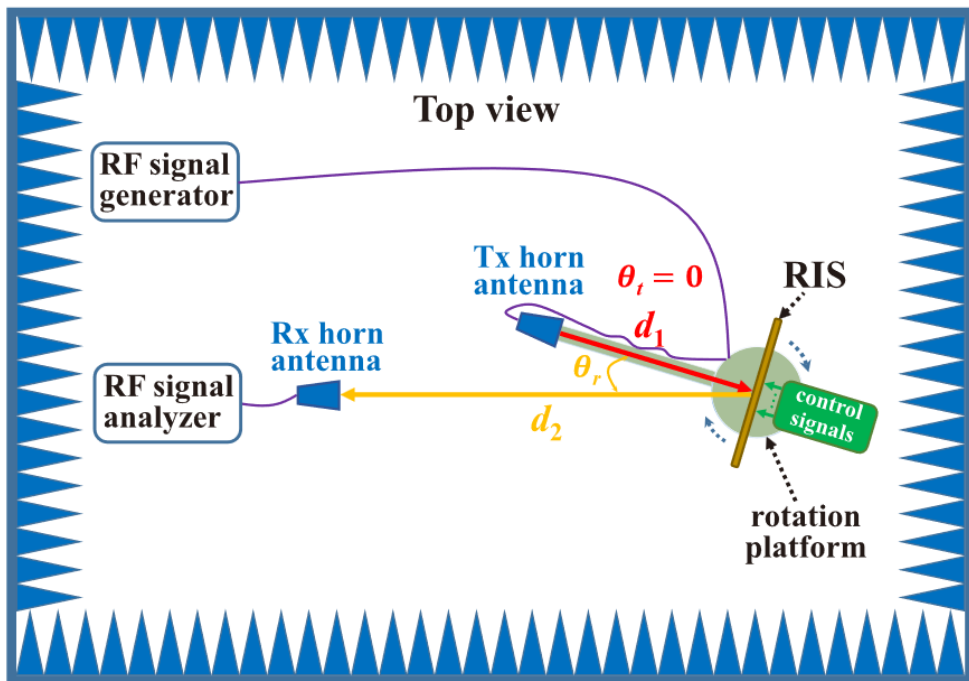
<sup>3</sup> WaveFidelity Inc., Taiwan



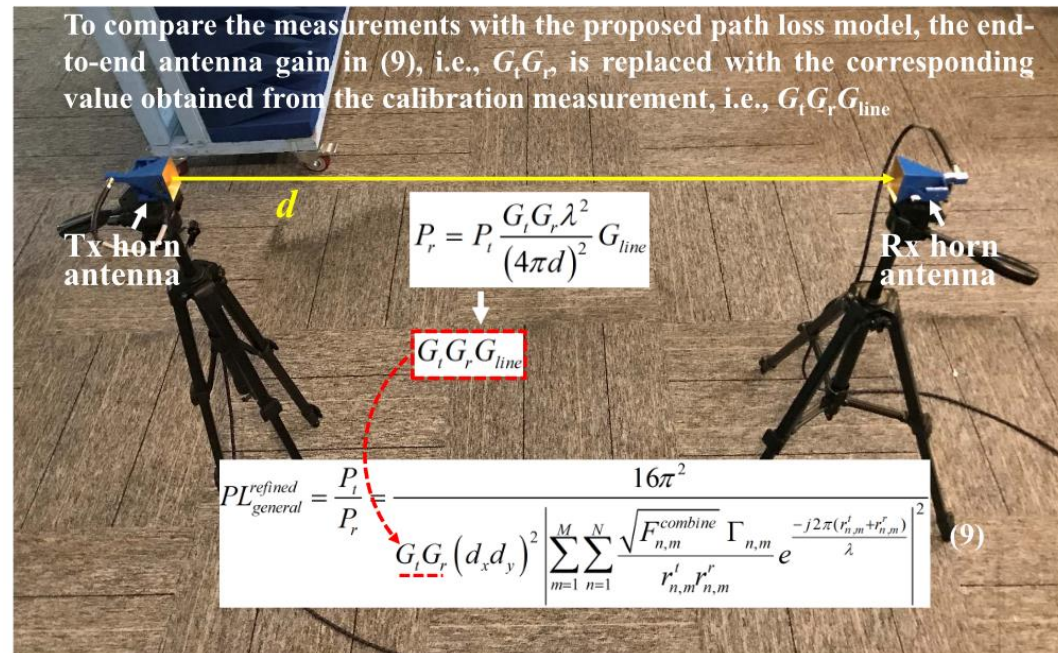
- CATR + 平面近場
- AoA/AoD涵蓋範圍小
- 平面近場轉換複雜



# 挑戰



- Bistatic (Planar wave\*2) → QZ大
- AoA/AoD有效角度



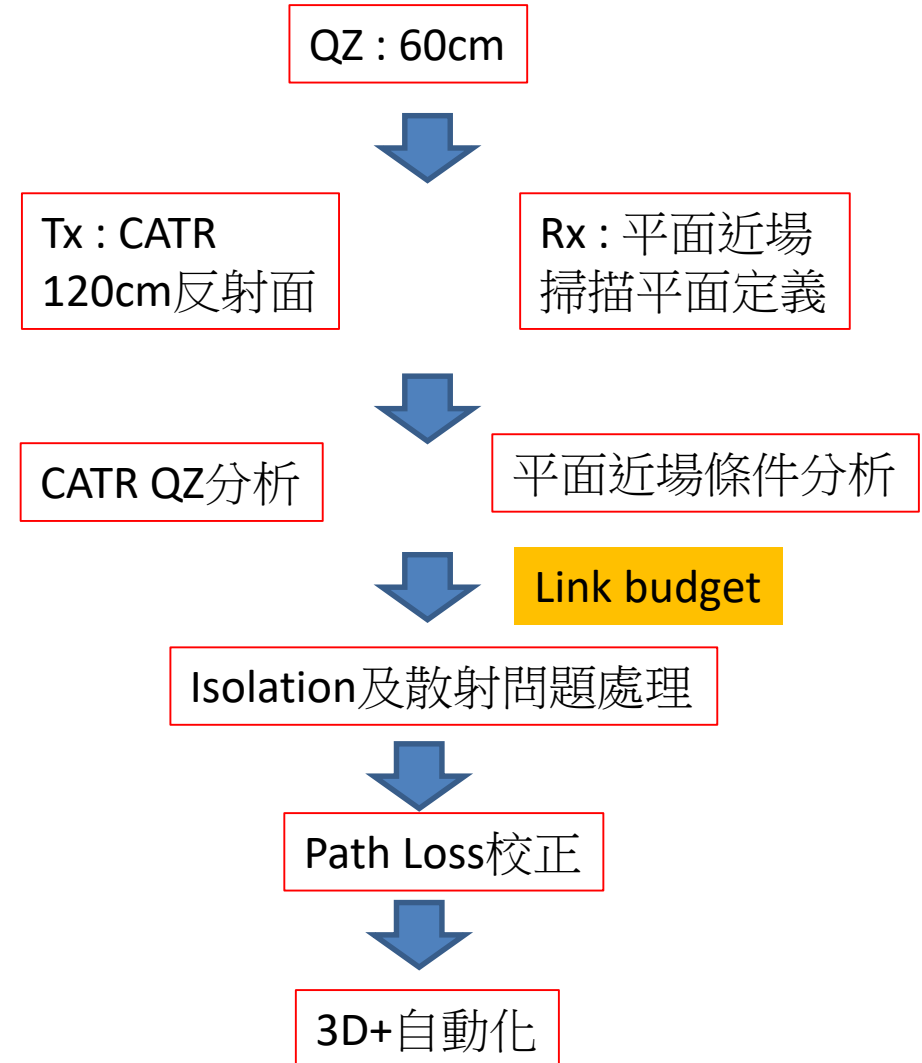
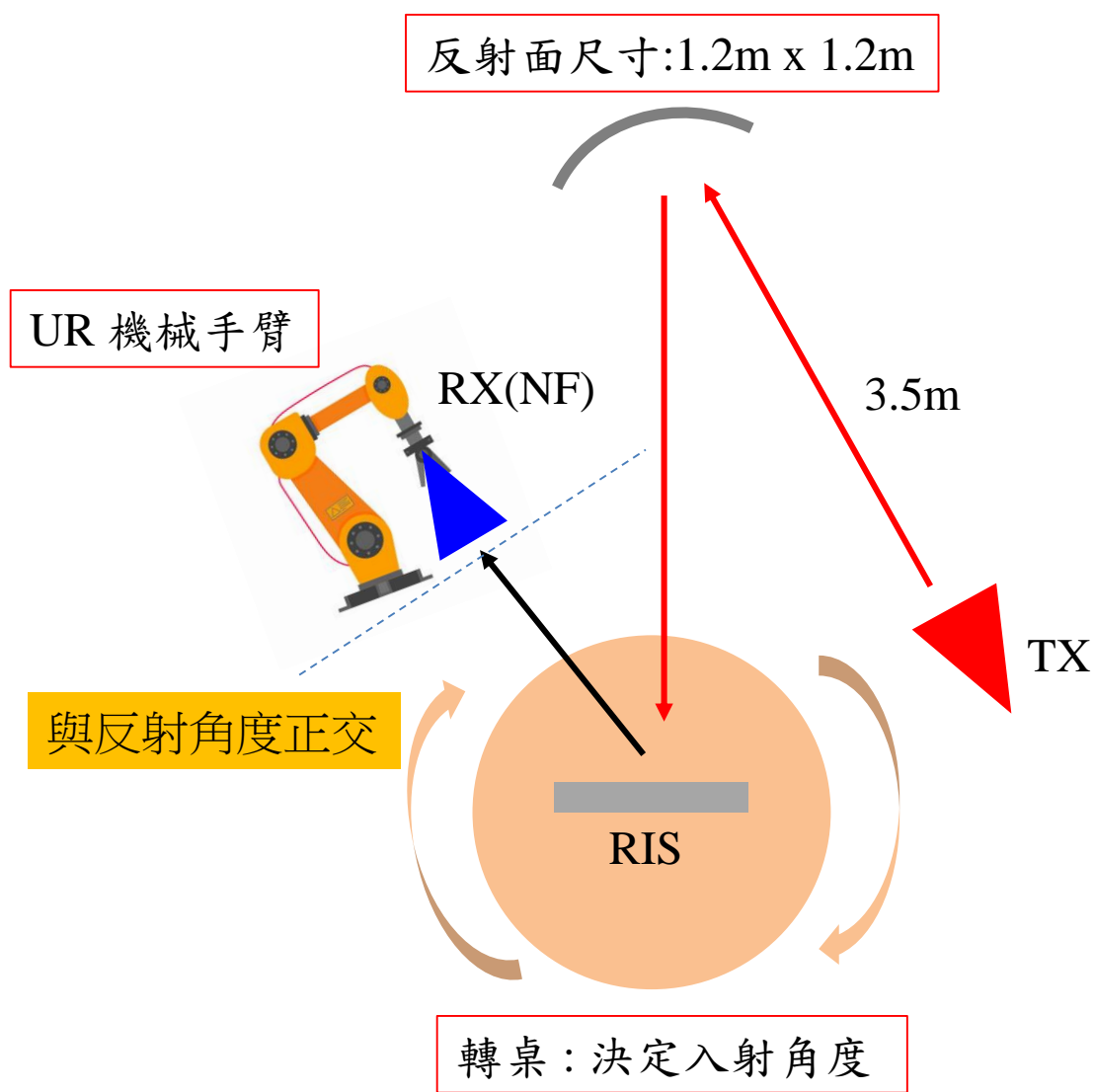
- Path Loss校正

# 大綱

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- RIS及OTA量測重要參數介紹
- RIS OTA量測文獻分享
- **RIS OTA量測系統介紹**
- 結論

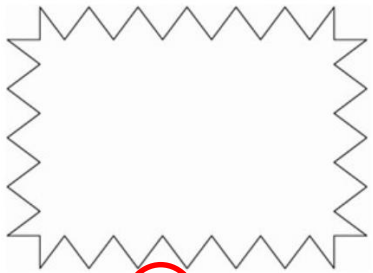
# RIS OTA量測系統設計



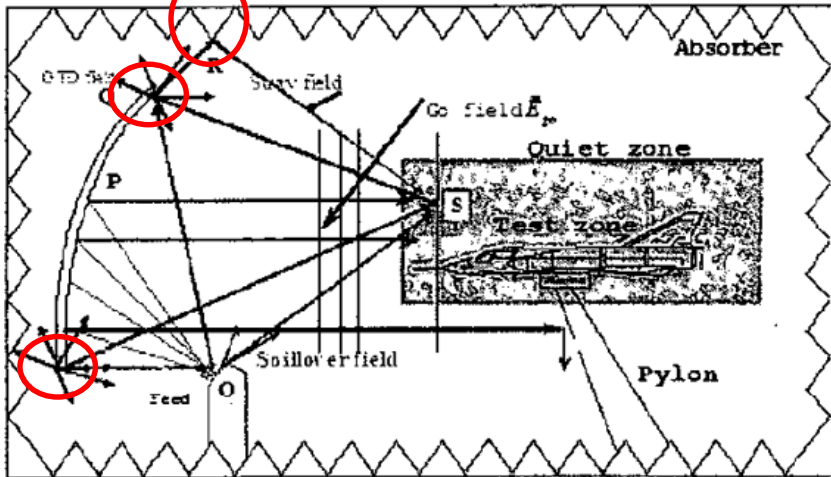
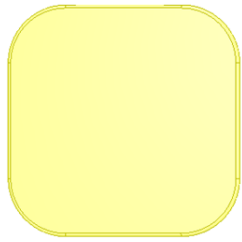
# QZ分析

## Edge treatment

Serrated Edge

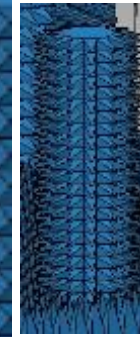
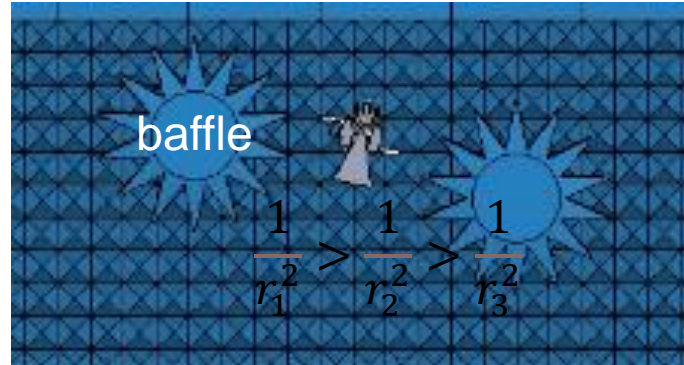


rolled edge



Chang, Dau-Chyrh et al. "Compact antenna test range without reflector edge treatment and RF anechoic chamber." IEEE Antennas and Propagation Magazine 46 (2004): 27-37.

## Unwanted radiation treatment

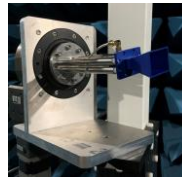


## Feed antenna

Wide band Stable phase center



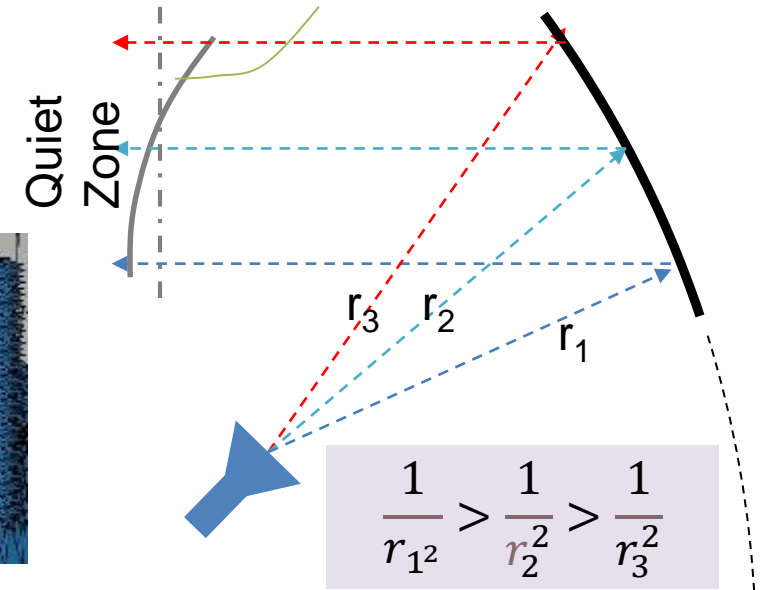
Rotational CP feed antenna



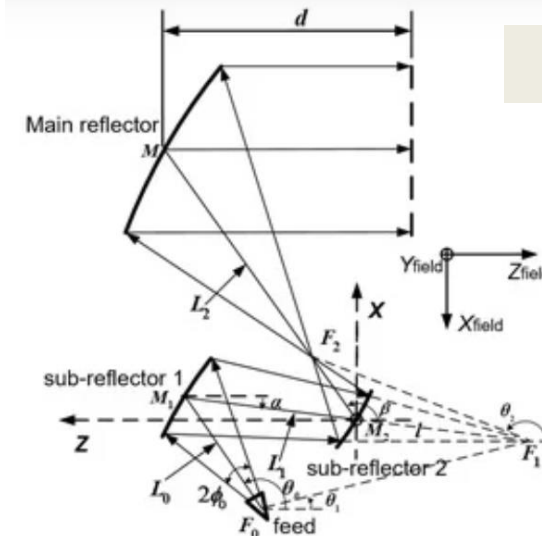
Low X-pol



## Amplitude taper

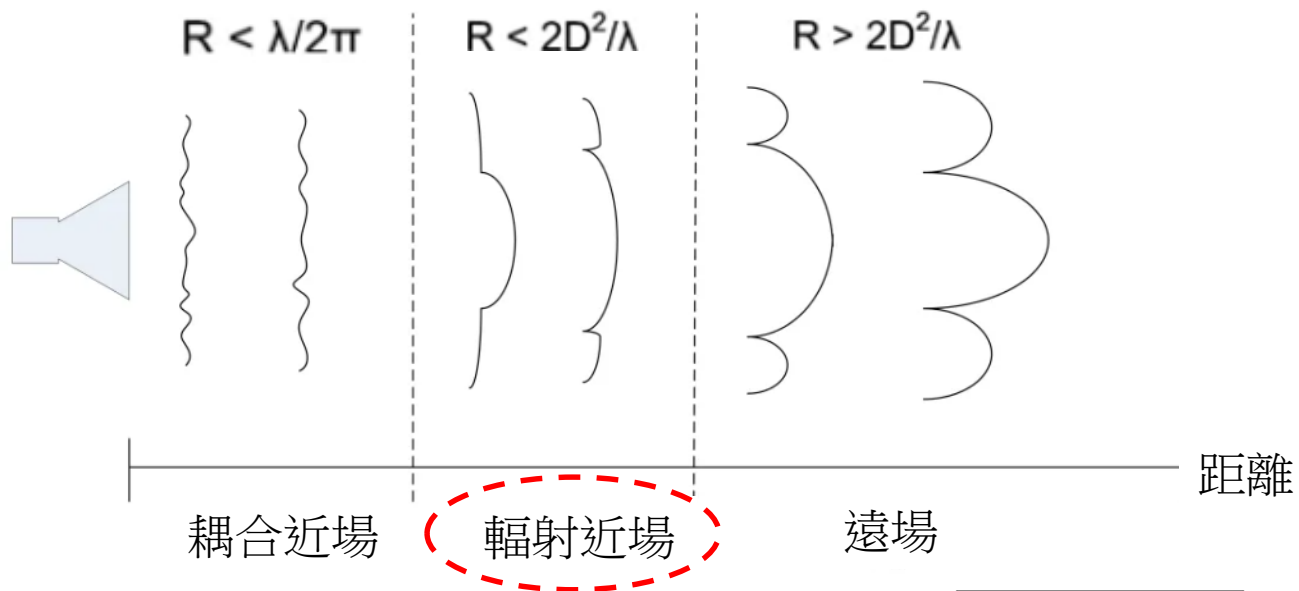


## Low X-pol

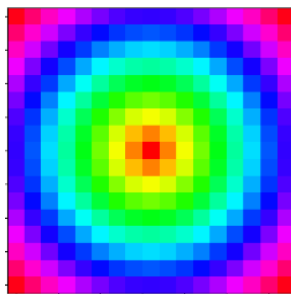


<https://doi.org/10.3390/electronics10141727>

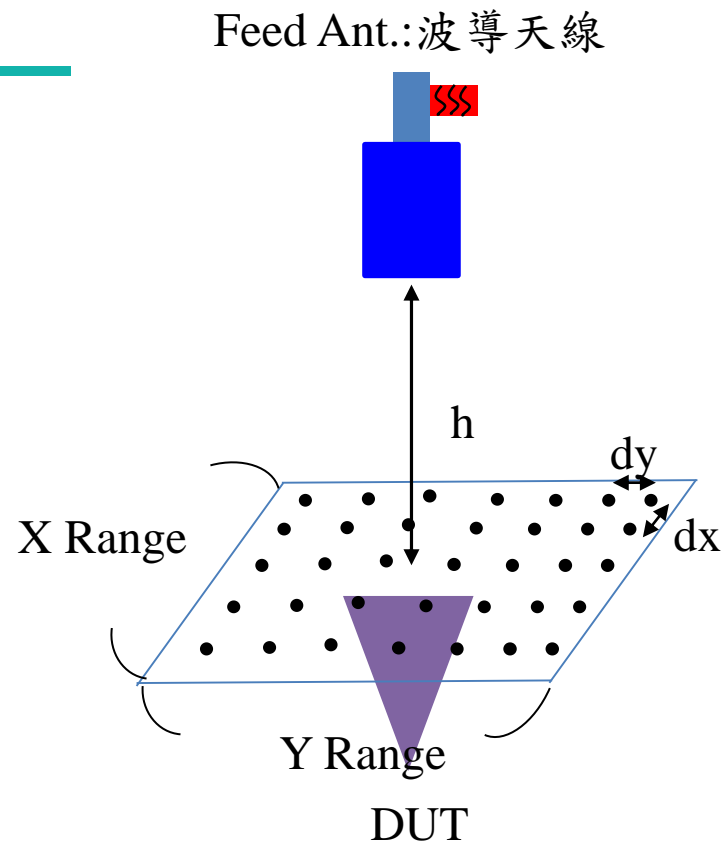
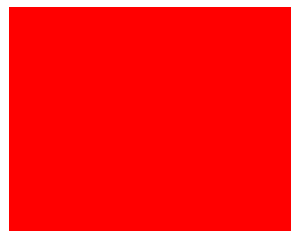
# 近場量測原則說明



近場量測: 輻射近場(球面波)



遠場量測: 遠場(平面波)



$dx/dy : < 1/20$ ; h 決定有效涵蓋角度/時間

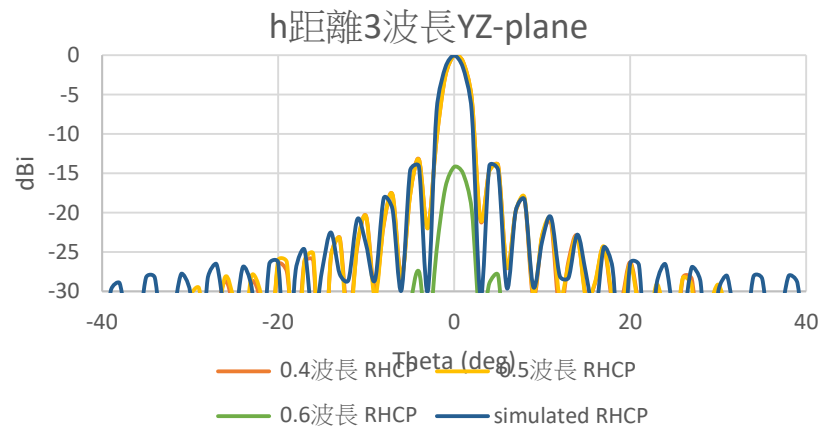
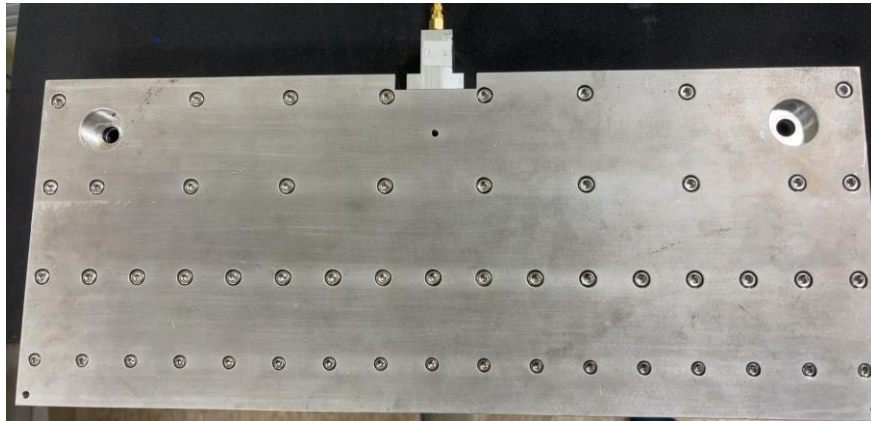
再利用程式進行轉換至遠場

Path Loss 校正

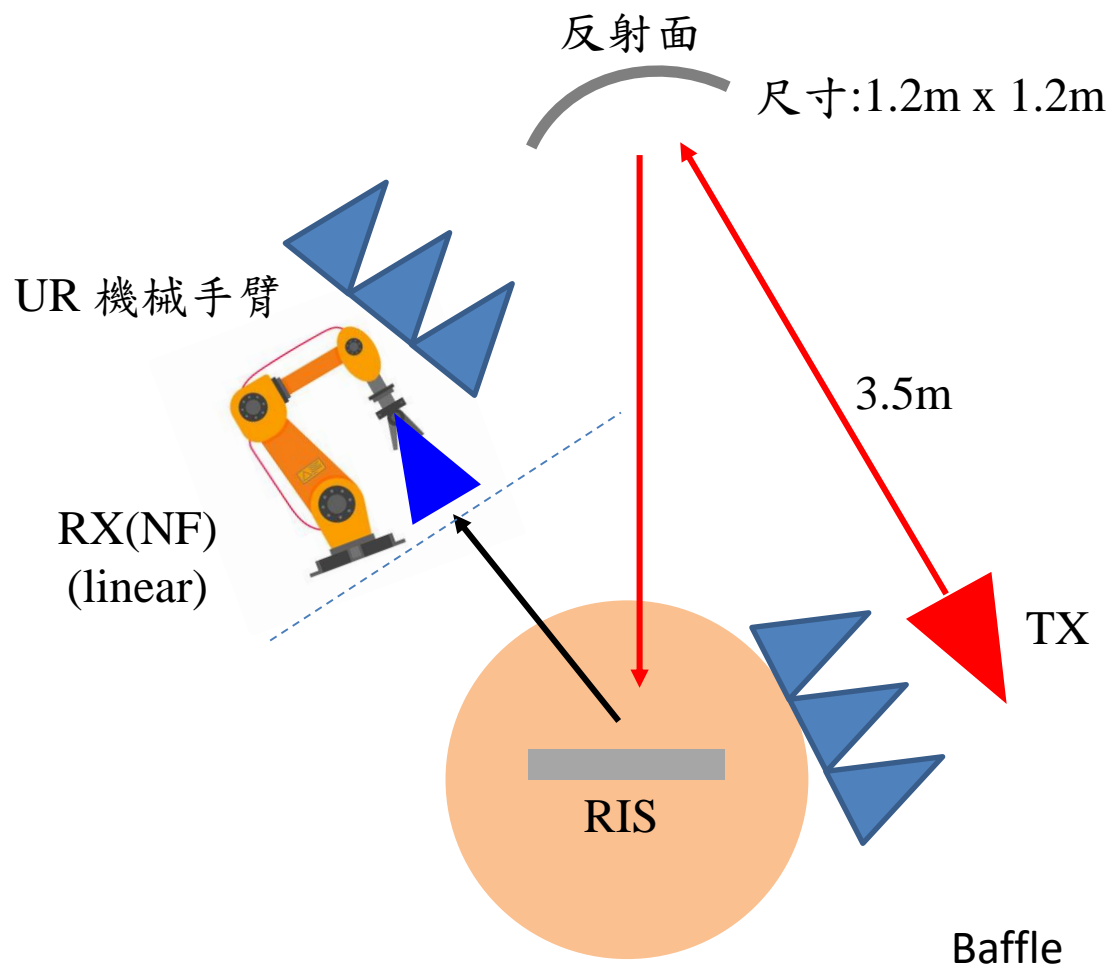
# NF系統&AUT說明

取樣間距: 2mm(約0.1波長)  
平面大小: 60cm X 60cm

DUT尺寸: 40cm x 2.5cm

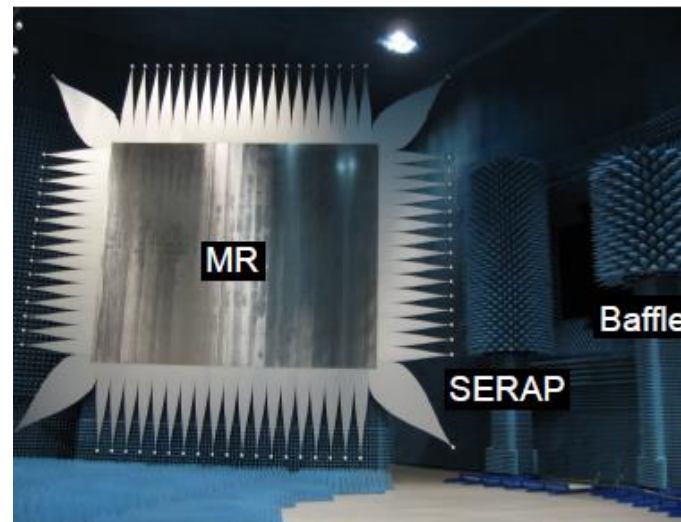
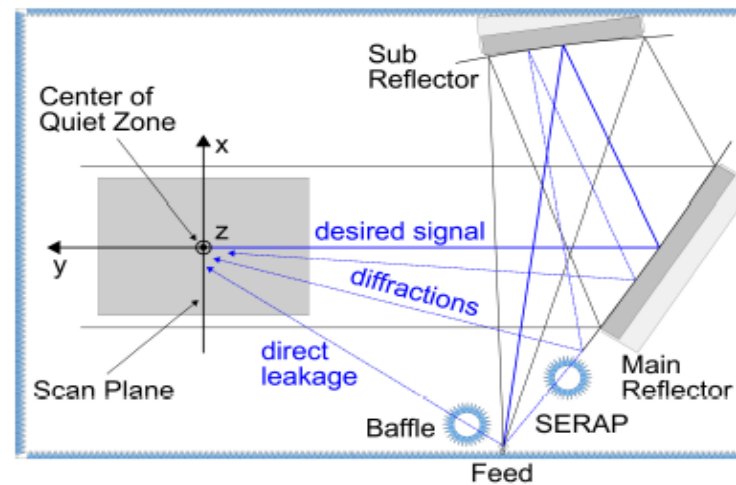


# Isolation & Scattering Effect issue



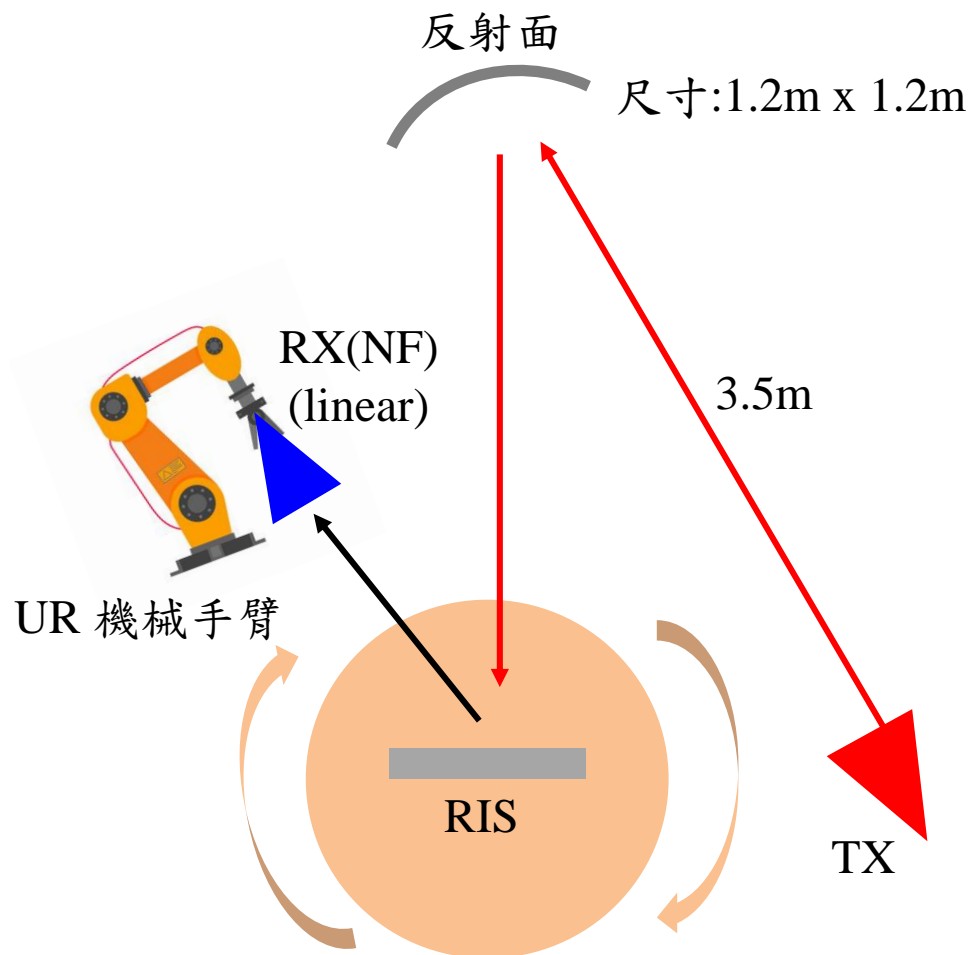
## Full-Wave Analysis of Compensated Compact Ranges Including Absorber Structures

T. M. Gemmer, C. H. Schmidt, A. Geise, J. Migl  
 Airbus Defence & Space, 81663 Munich, Germany, thomas\_gemmer@gmx.de, carsten.schmidt@airbus.com

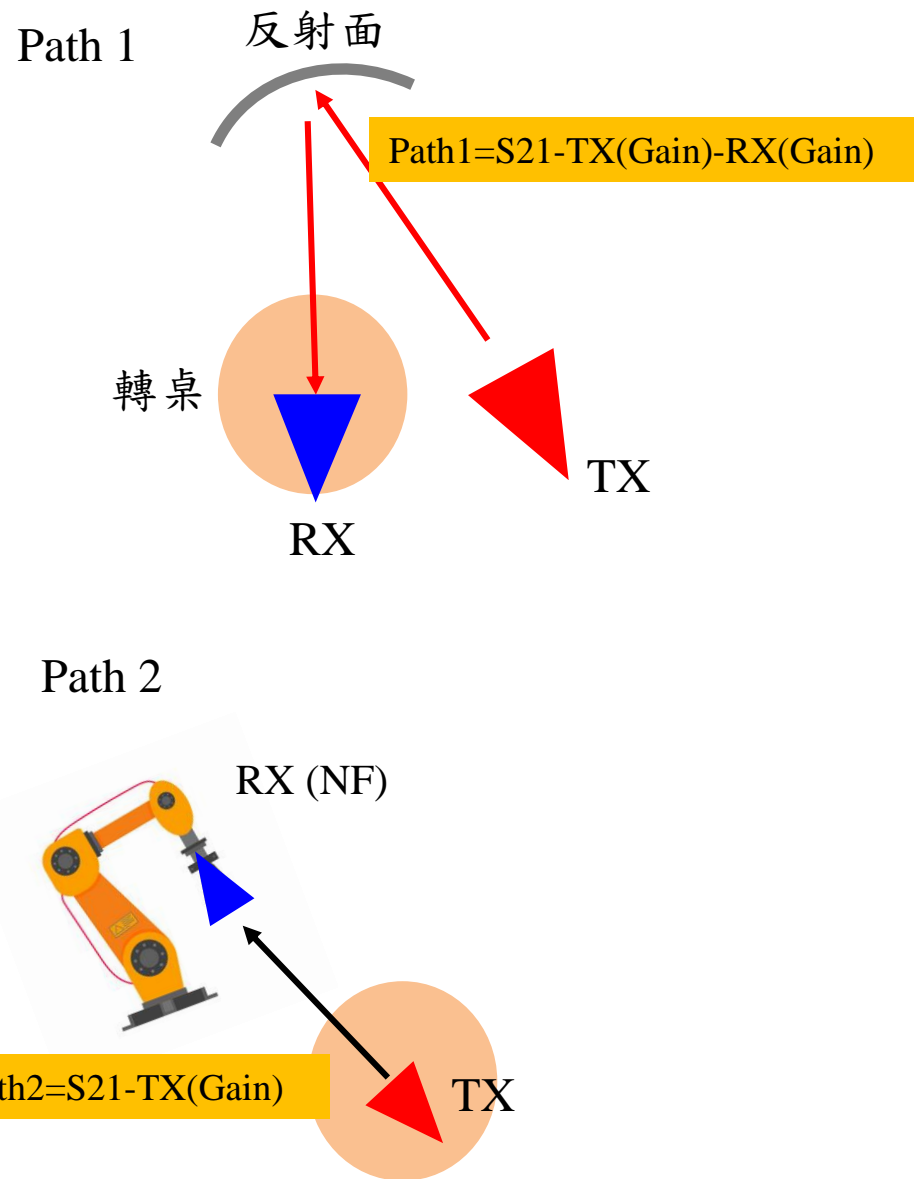


# Path Loss校正-SGH分段

使用縮距遠場+平面近場

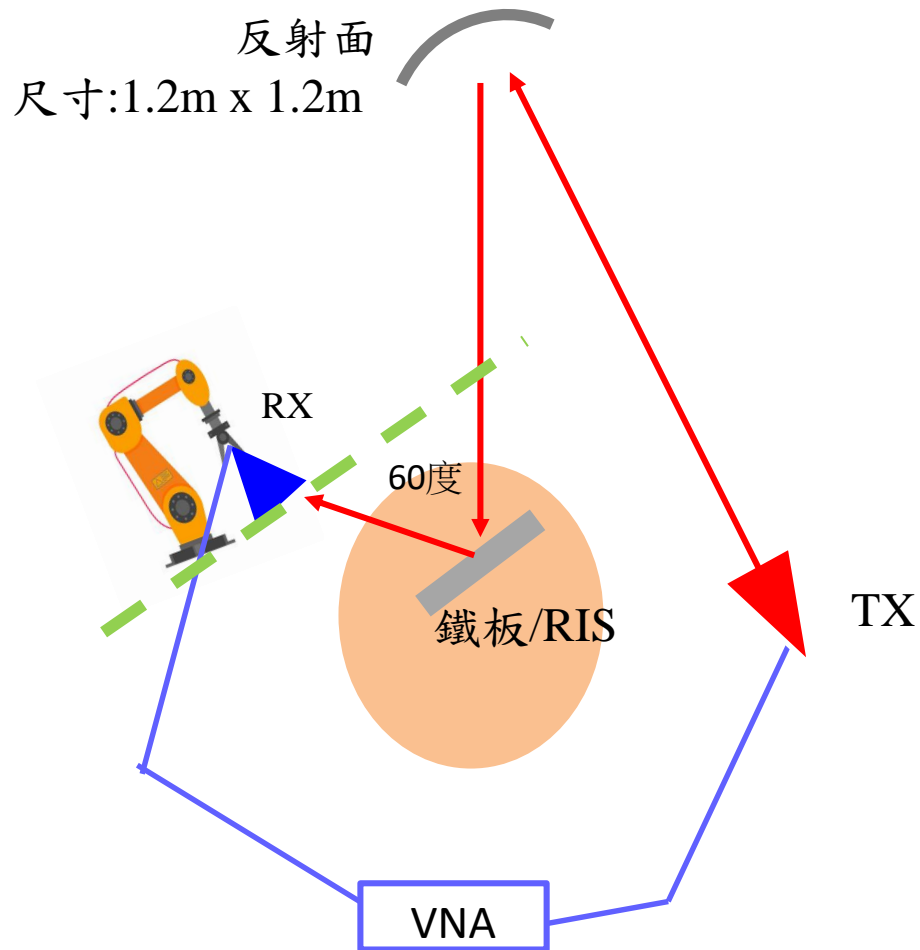


轉桌：決定入射角度





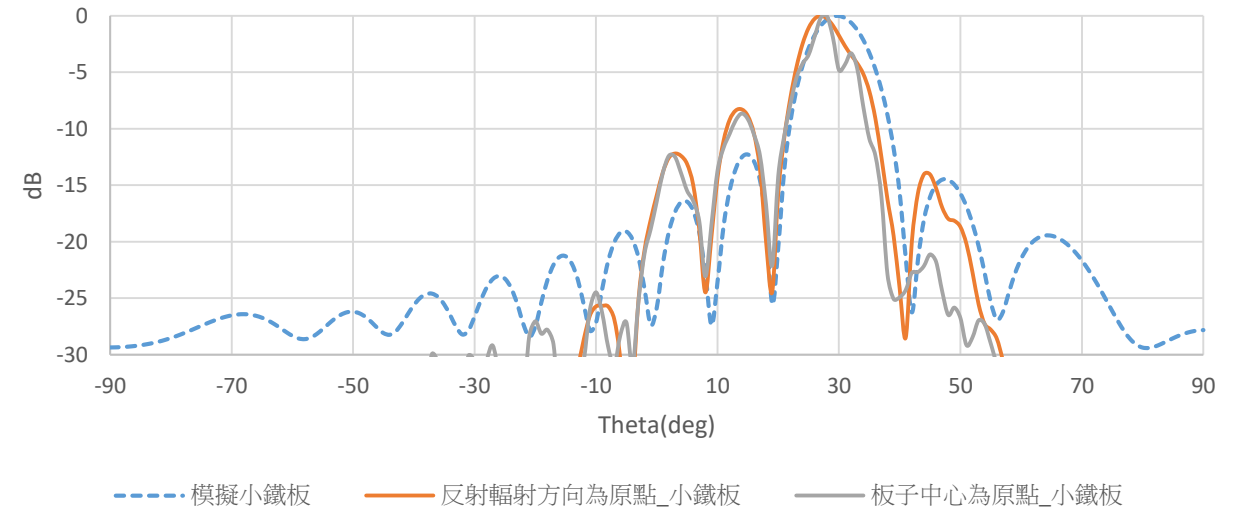
# Path Loss校正-替代法



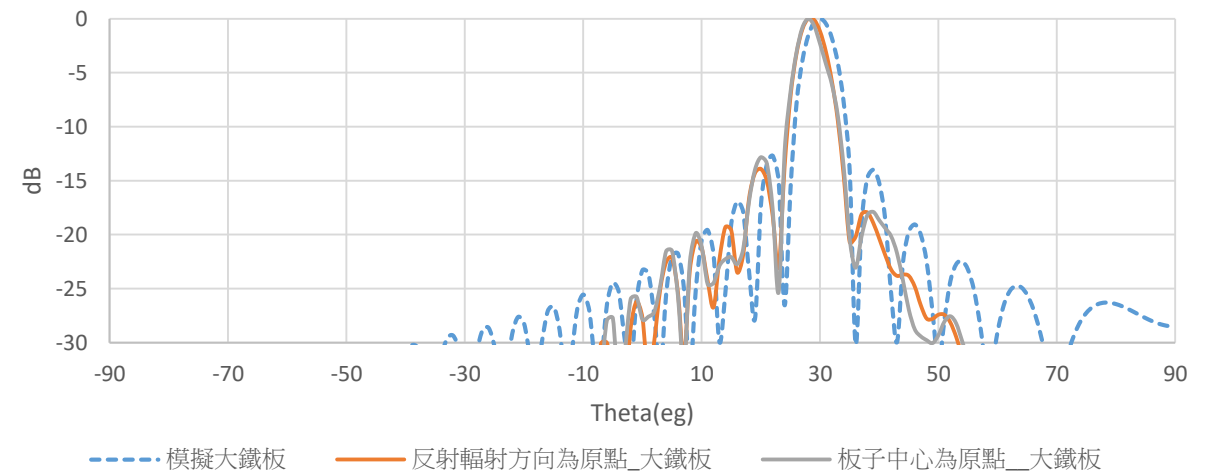
以反射輻射方向為原點進行量測 → 有效角度

不同AoA/AoD Aperture size不同

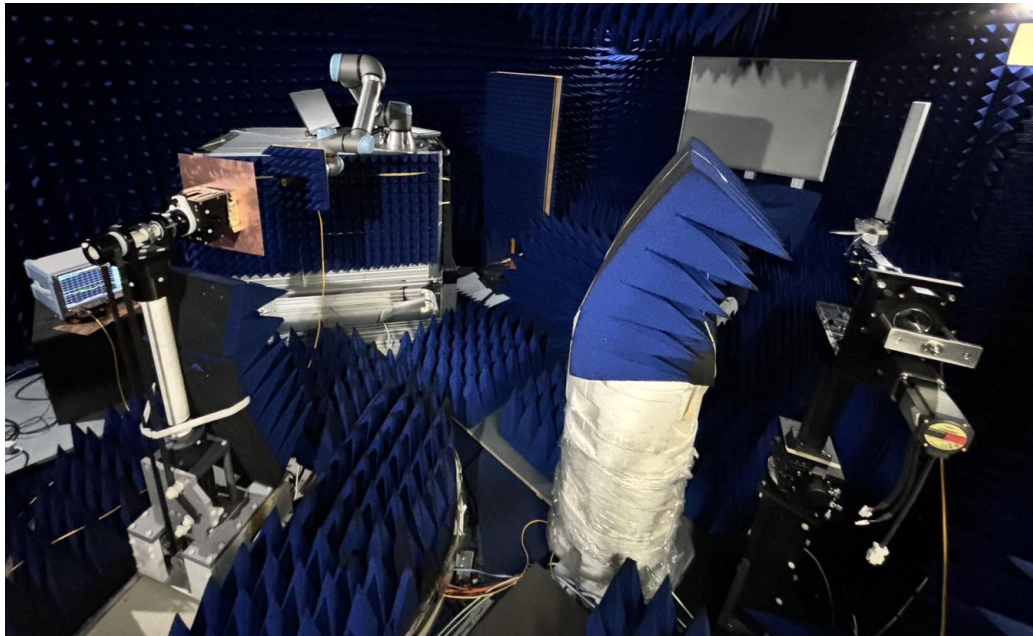
小鐵板場型比較



大鐵板場型比較



# 鐵板Path Loss差異



Path Loss 為量測值-模擬Gain  
三種鐵板之間最大差異為1.1 dB

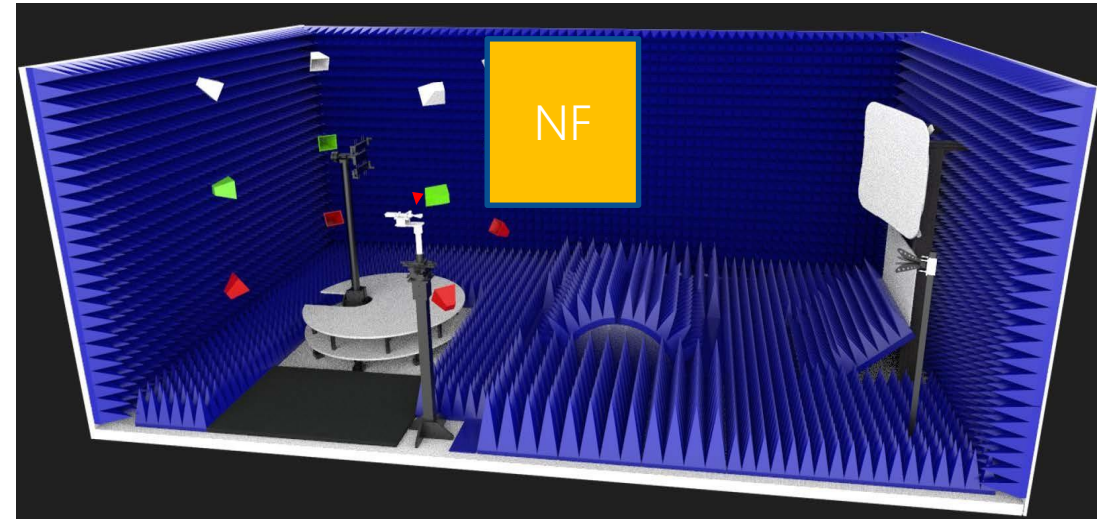
差異來源我認為是QZ不理想原因較大  
因為發射天線並非為設計過的CATR發射天線  
所以還是需要看待測物大小變更鐵板大小

	28 * 28 cm	20 * 20 cm	15 * 15 cm		
反射輻射方向為中心 (量測值)	-65.2 dB	5.0	-70.2 dB	5.3	-75.5 dB
模擬gain	18.6 dBi	6.1	12.5 dBi	4.9	7.6 dBi
Path Loss	-83.8 dB	-82.7 dB	-83.1 dB		

# 架構尺寸



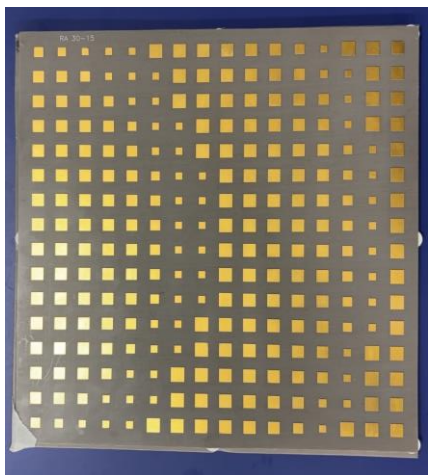
System specification (typical)	
External Dimension	7 × 4 × 3 m <sup>3</sup> (L × W × H)
Reflector Focal length	350 cm
Isolation	> 90 dB (0.6GHz~80GHz)
Absorber	Pyramid type absorber
Reflector specification (typical)	
Frequency range	0.6-80 GHz
Size	120 × 120 cm <sup>2</sup>
Edge treatment	Rolled edge
Quiet zone size	Diameter 60 cm
Cross polarization isolation	> 30 dB
Amplitude taper	3 dB
Amplitude ripple	±1.5 dB



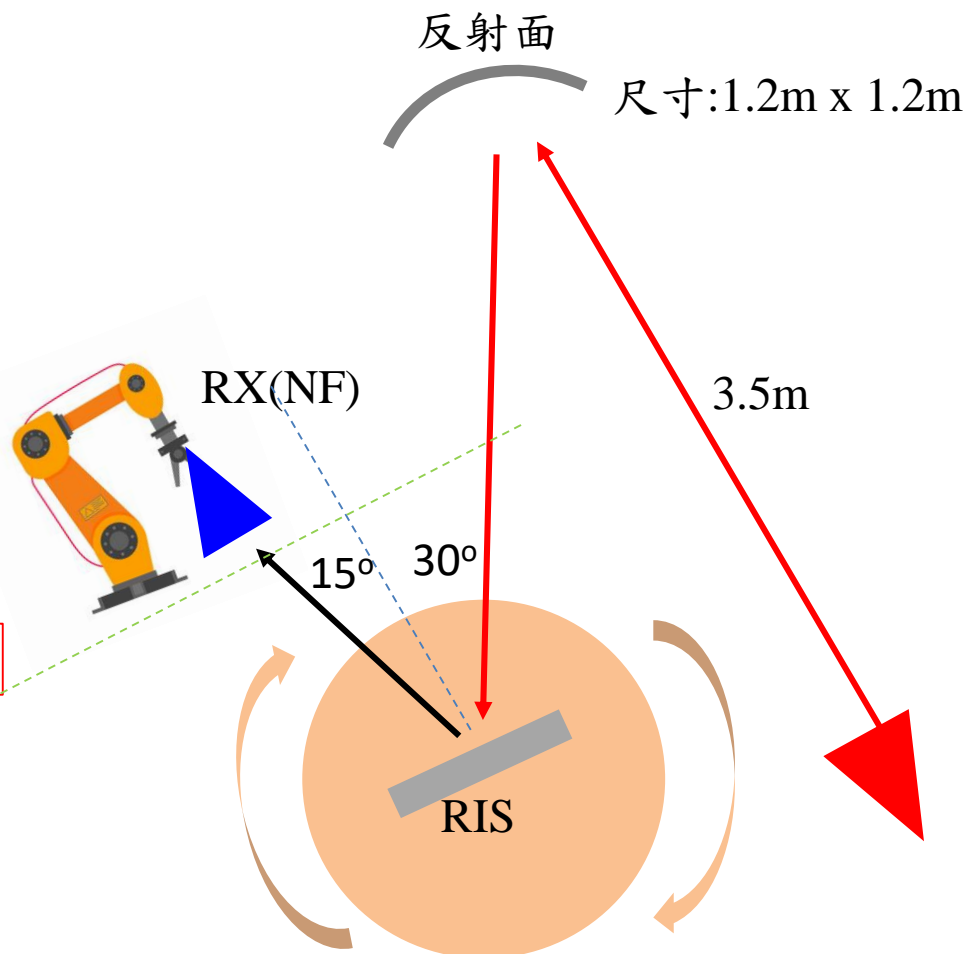
# 模擬及量測結果

差異可能原因 -  
 Peak角度：RIS法向量定位  
 HPBW：QZ&RIS設計相位

20 \* 20 cm



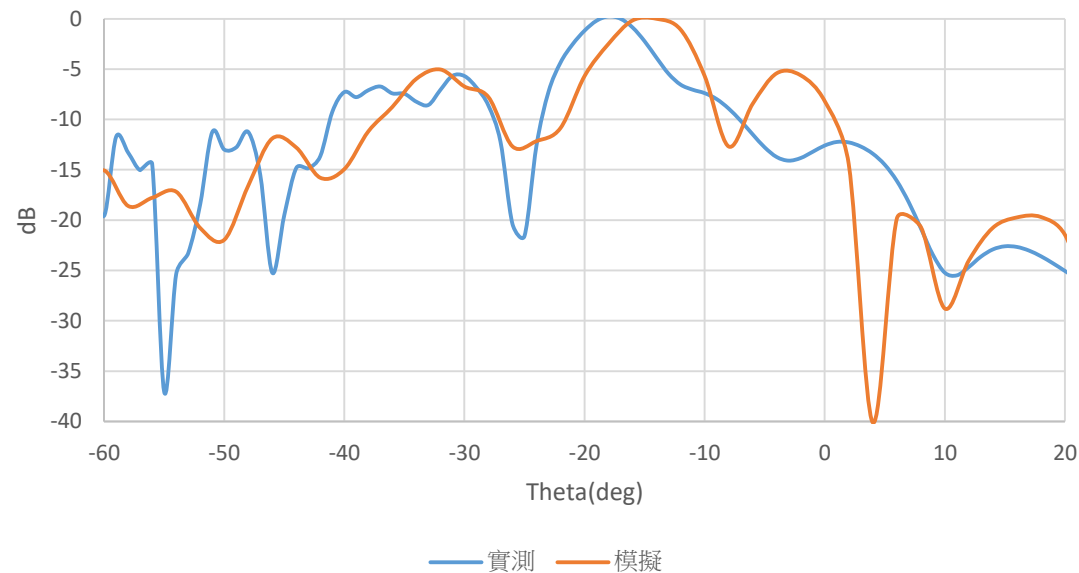
入射:30度；反射:15度



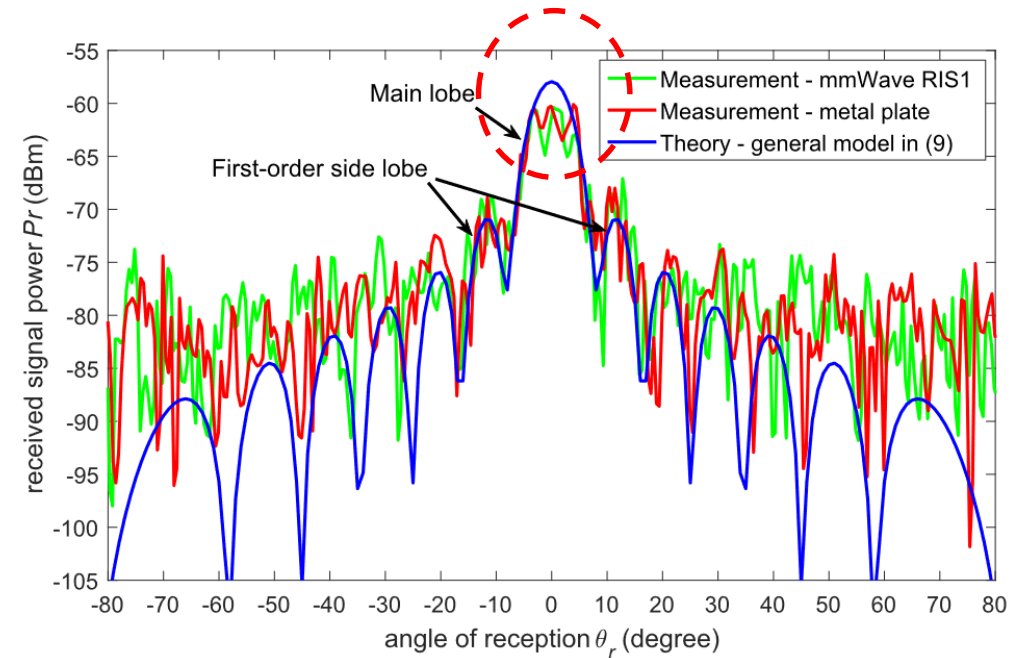
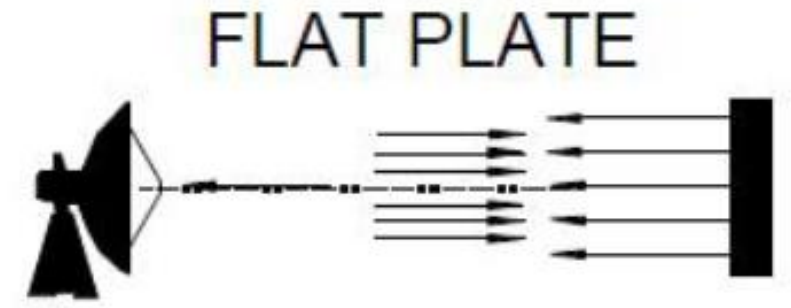
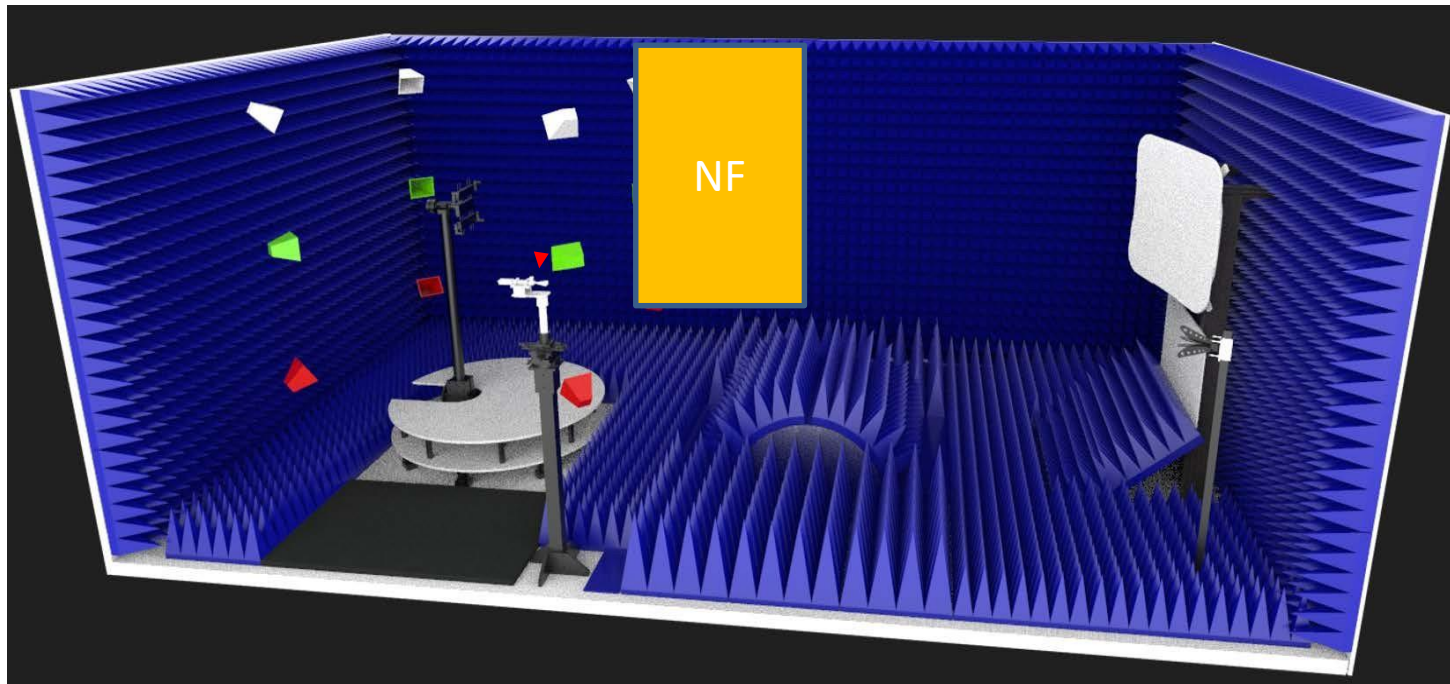
	模擬	量測	差異
Peak 角度	15(30)	17(32)	2(2)
HPBW	8(8)	4(6)	4(2)

( )為同尺寸鐵板結果

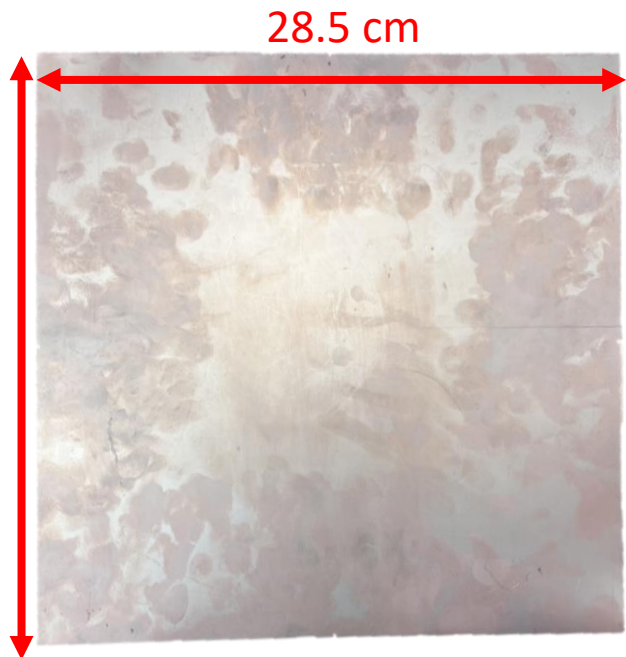
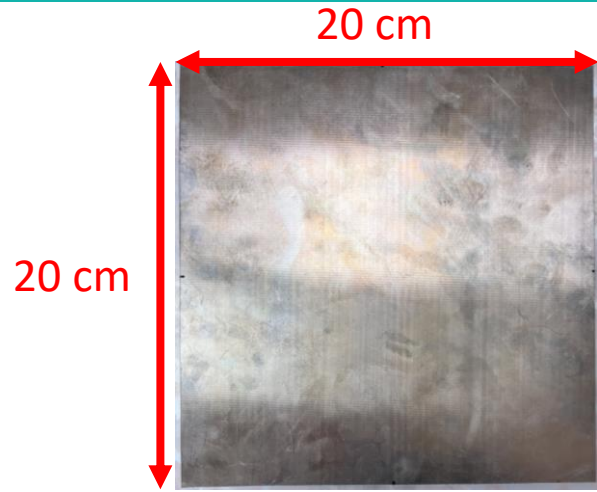
RIS模擬實測場型圖



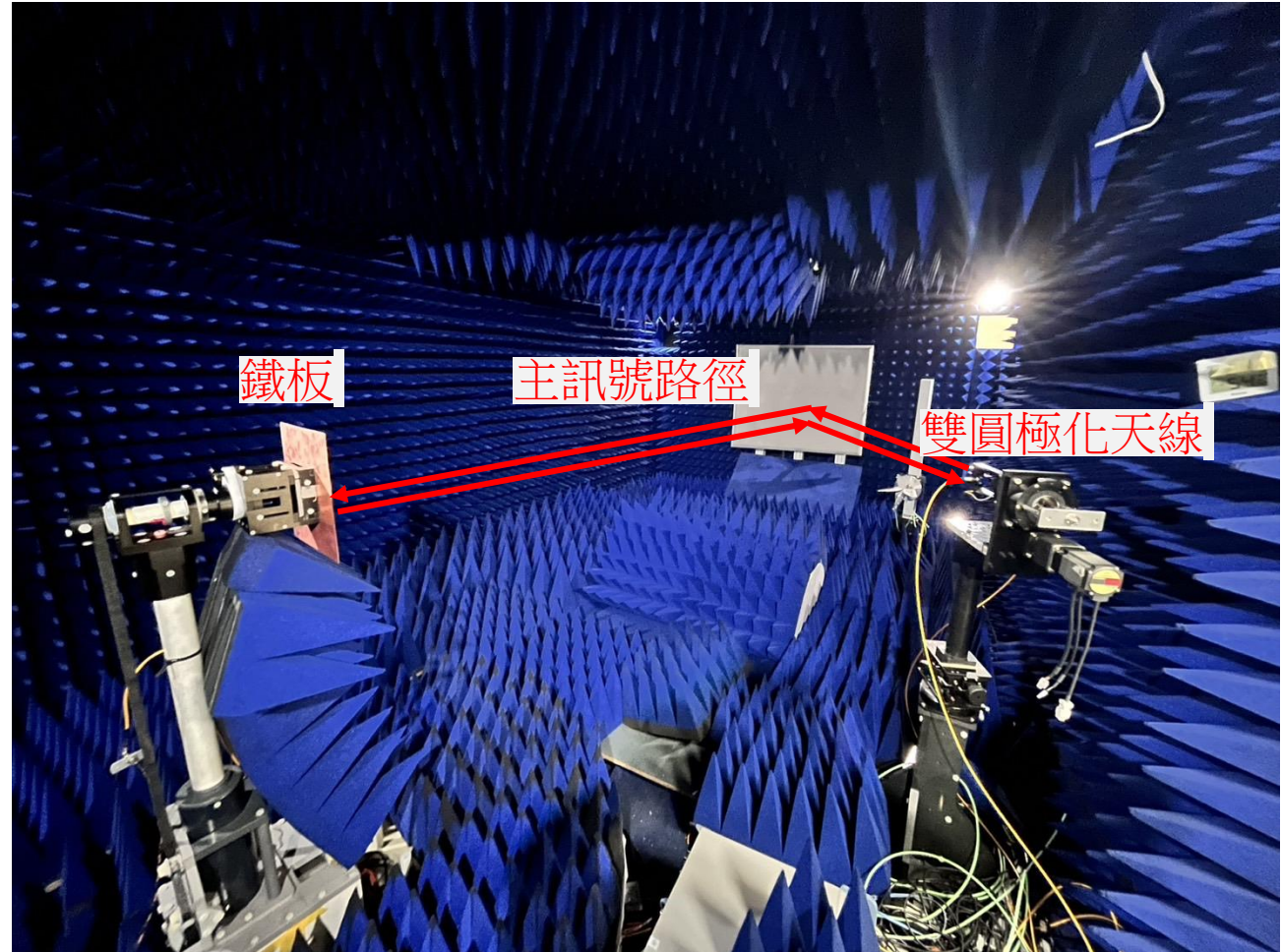
# AoA = AoD條件量測?



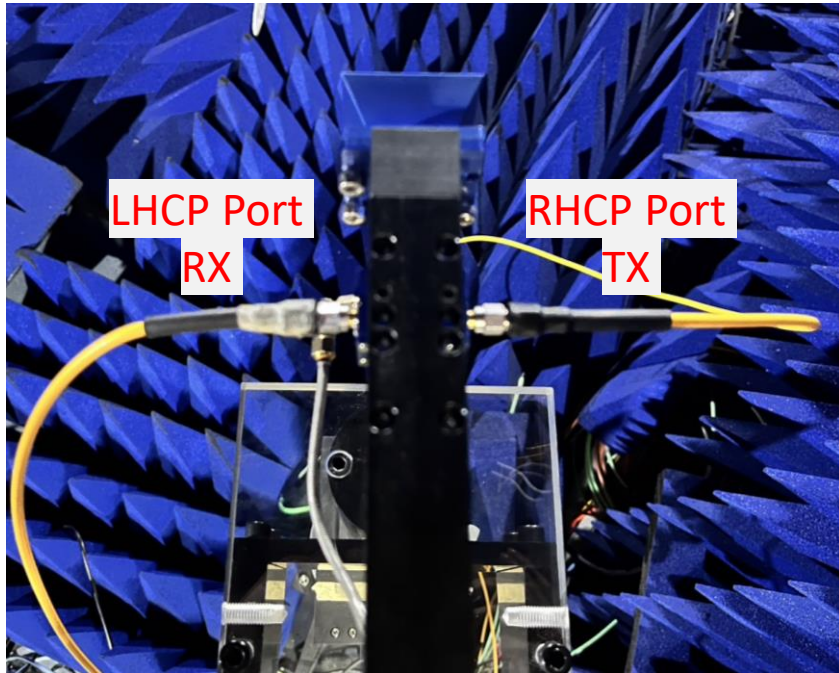
# 雙圓極化量測



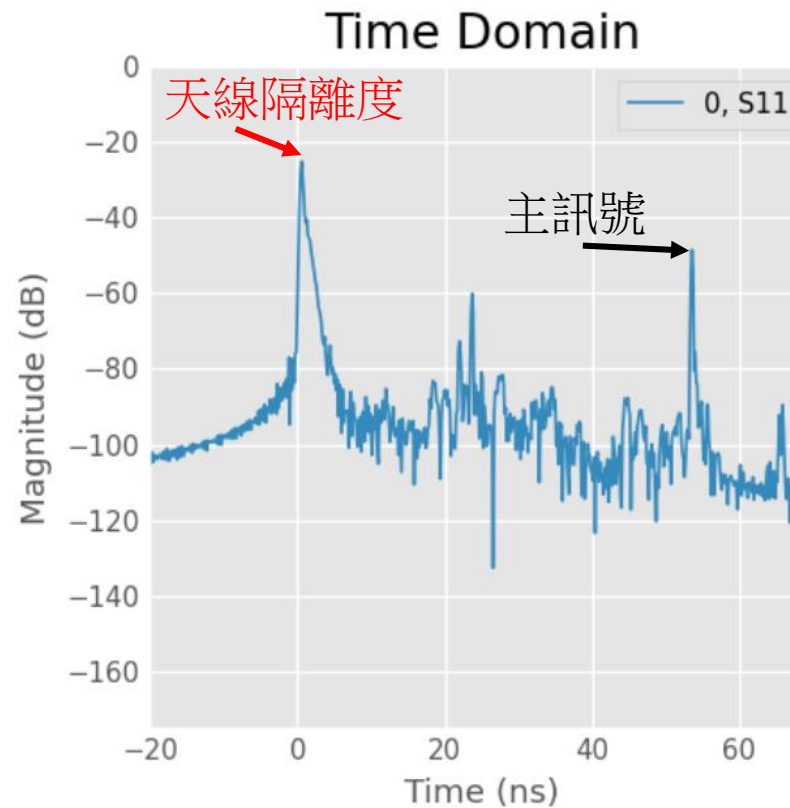
量測環境



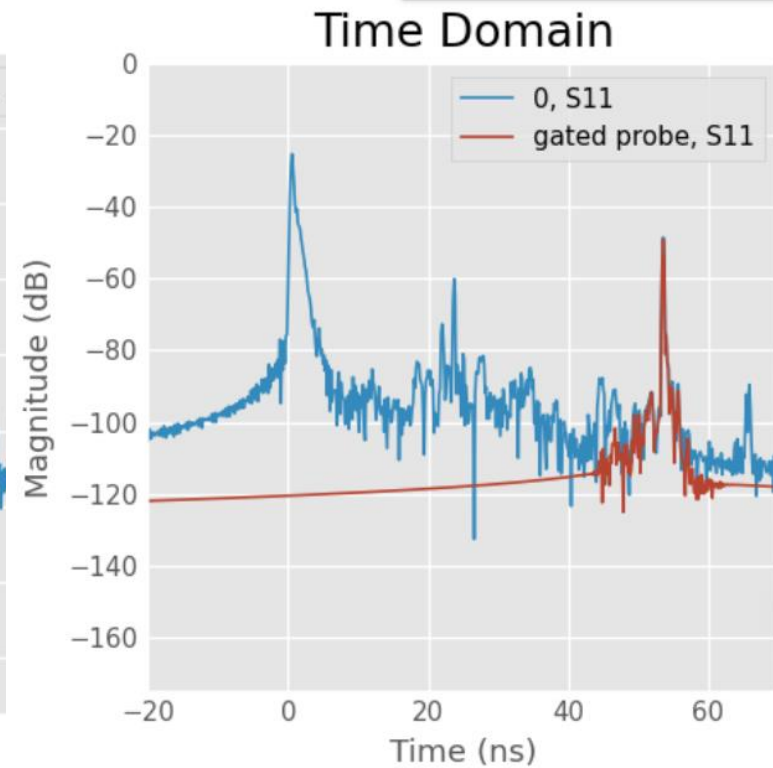
# Time Gating



No Time gating

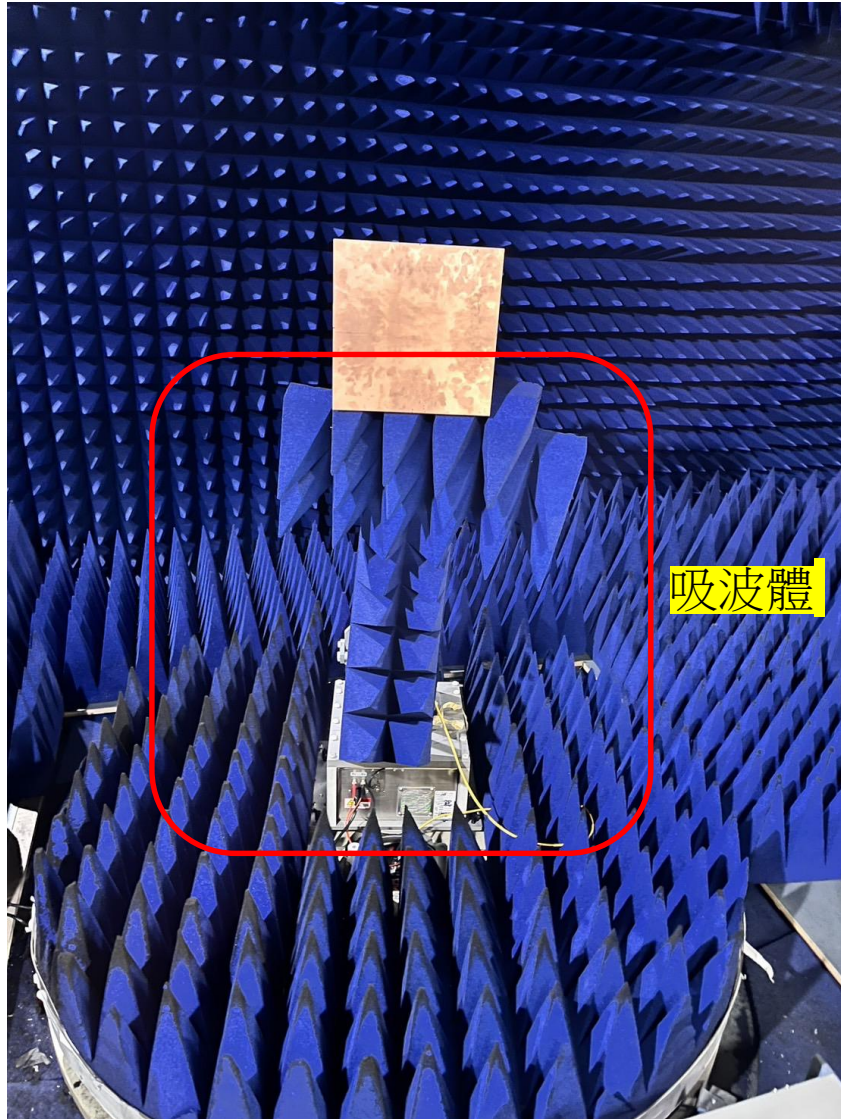


Time gating



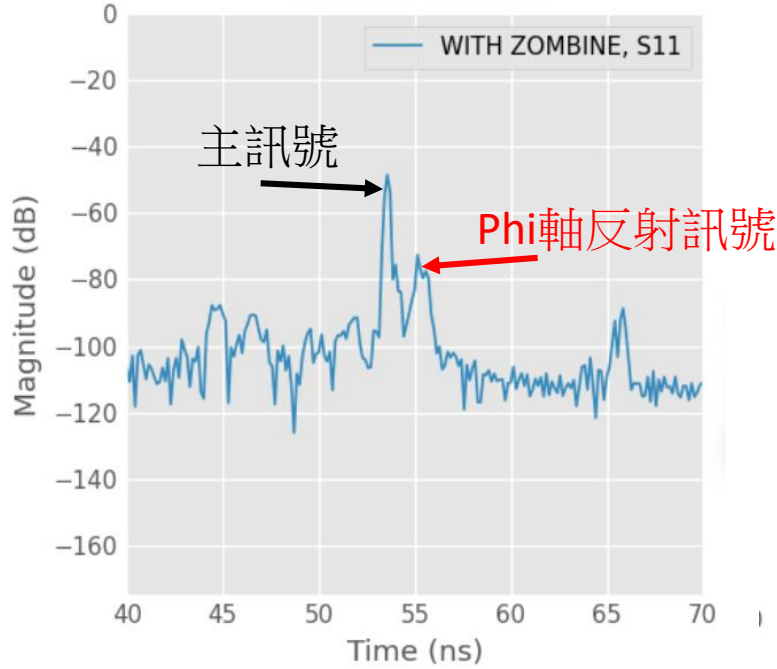
因接收與發射都在同一支天線上，在網儀上會量測到天線隔離度，所以必須用 Time gate 來消除，Time gate 的範圍取主訊號左右雜訊牆最低處。

# 治具散射分析



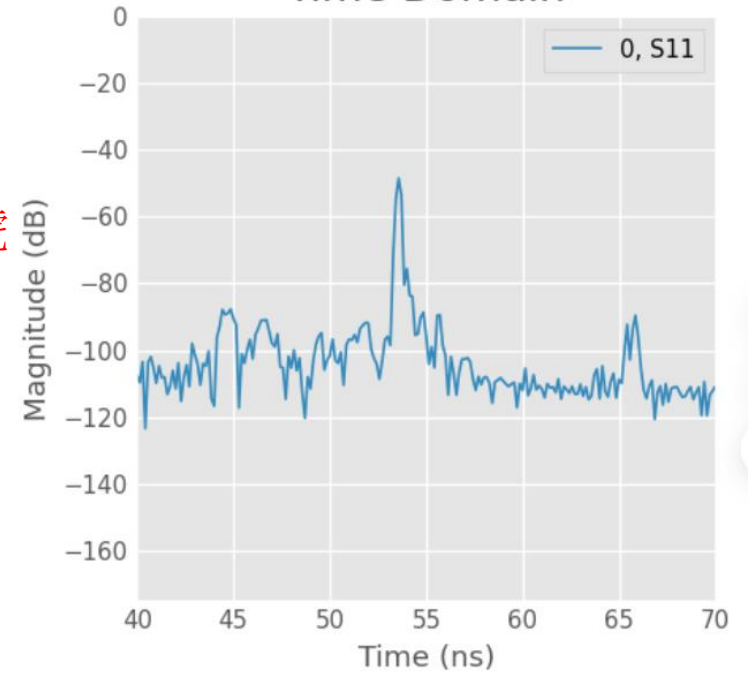
沒有吸波體

Time Domain



有吸波體

Time Domain



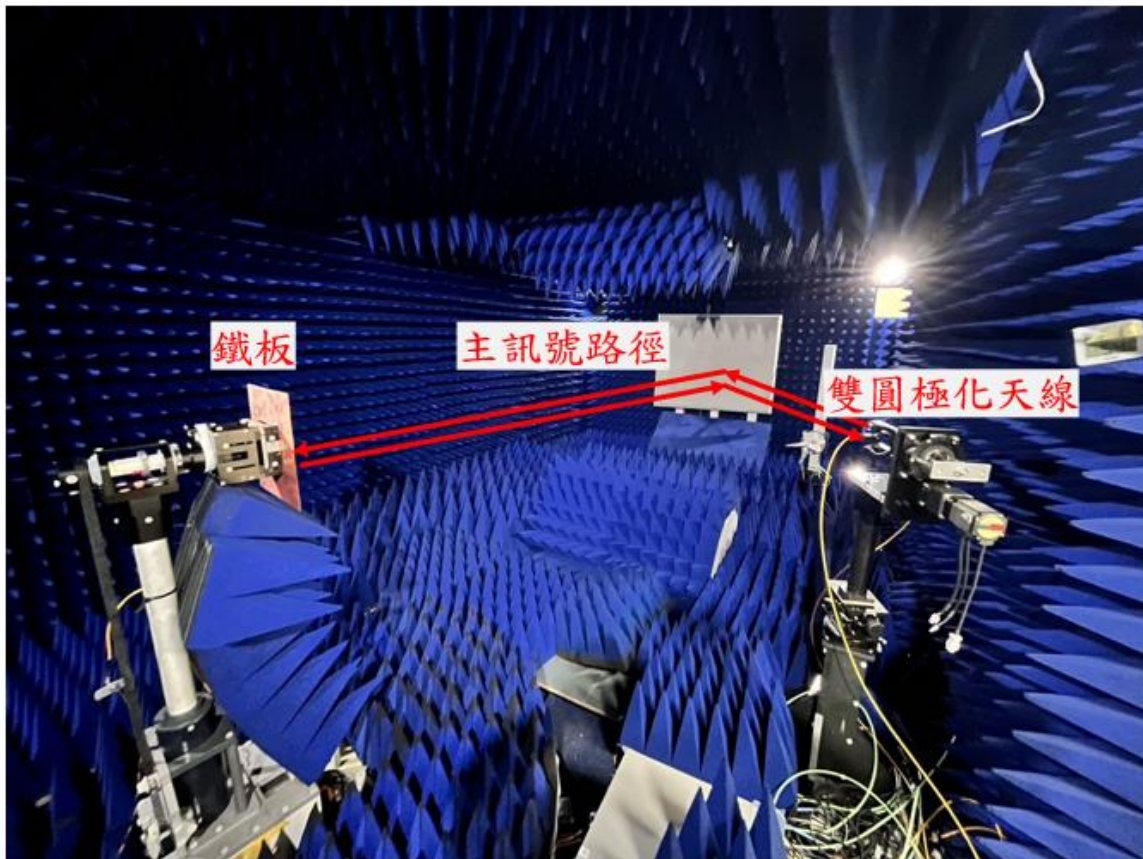
在左上時域圖中，可以發現圖中有兩個較強的訊號，而反射訊號離主訊號太近故無法用time gate 消除，故在Phi軸手臂放滿吸波體。並且反射訊號消失。



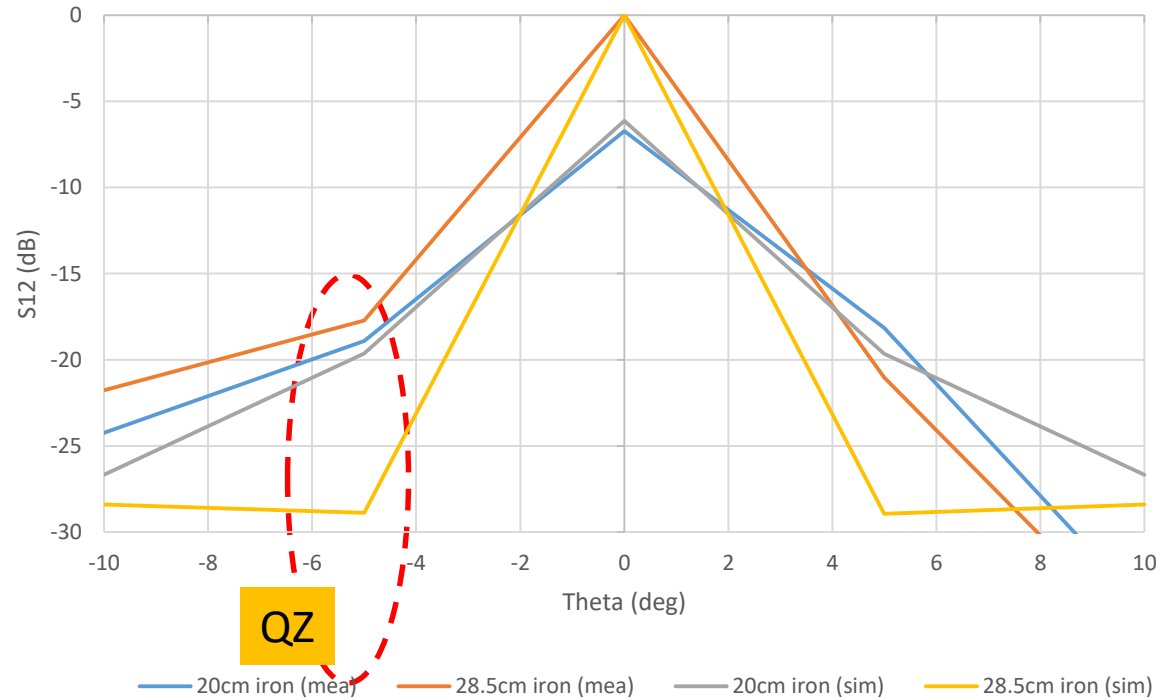
# 量測結果

	RCS計算值	RCS模擬值
28.5cm iron	21.1 dBsm	21 dBsm
20cm iron	15 dBsm	14.9 dBsm

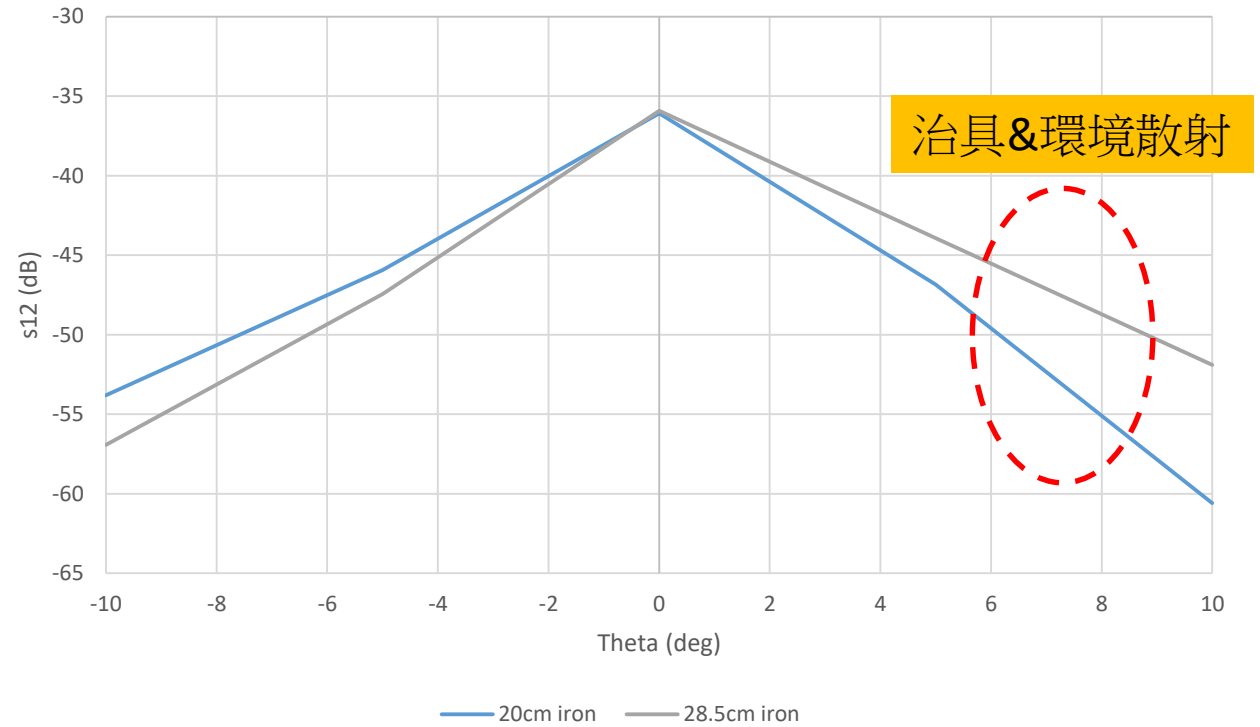
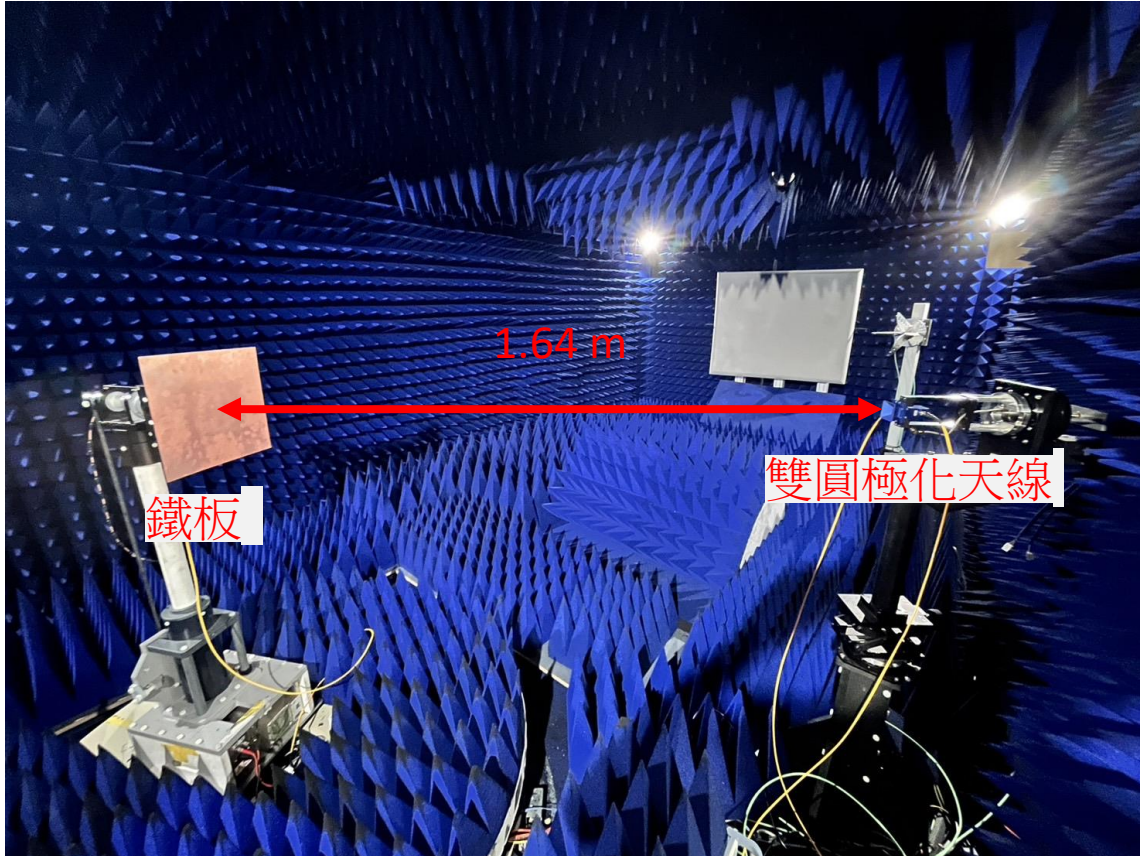
$$\sigma \text{ (dBsm)} = 10 \log_{10} [\sigma \text{ (m}^2\text{)}]$$



正規化後數值比較



# 直接遠場??



QZ不足：  
20 cm 鐵板、28.5 cm 鐵板量測結果相近

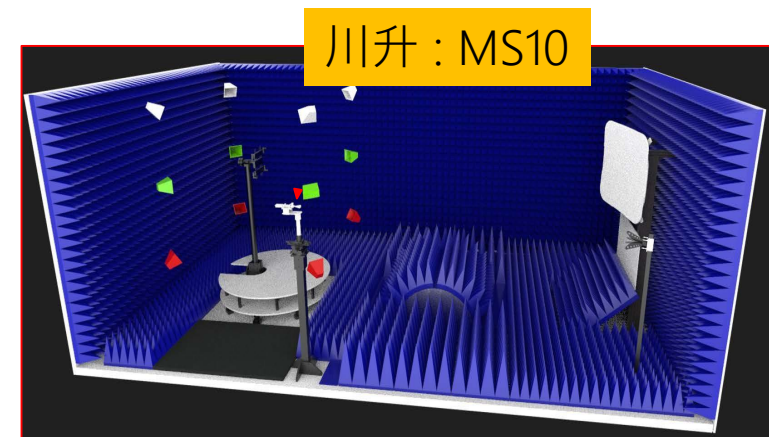
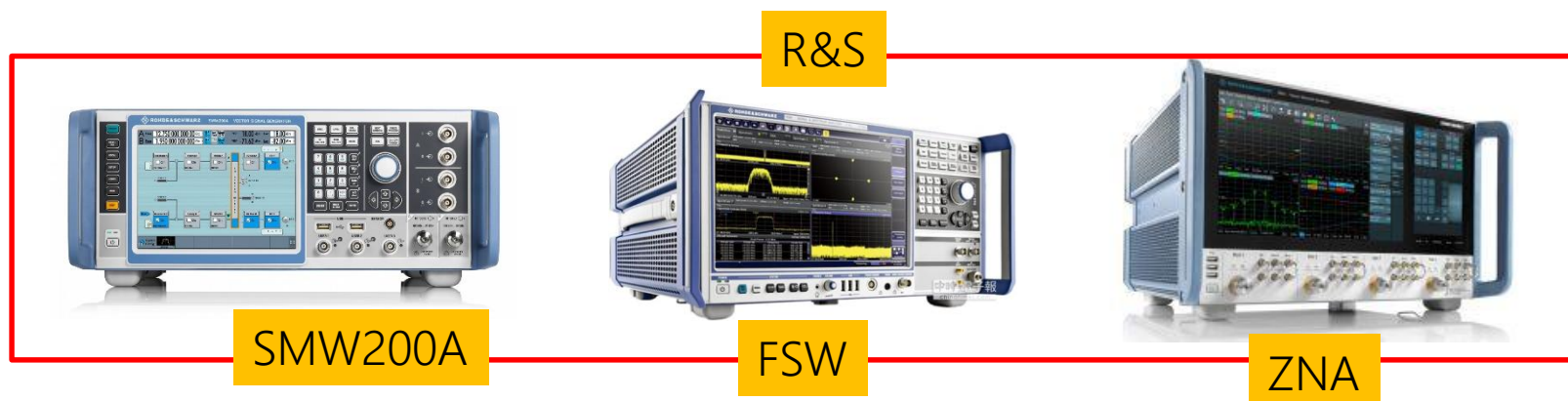
# 大綱

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- RIS及OTA量測重要參數介紹
- RIS OTA量測文獻分享
- RIS OTA量測系統介紹
- **結論**

# 結論 – RIS OTA

- ✓ RIS OTA挑戰：
  - ➔ Bistatic、Path loss校正、AoA/AoD有效角度、散射
- ✓ CATR Tx + NF Rx ➔ Bistatic
- ✓ 替代法 ➔ 解決path loss校正
- ✓ 轉台調整入射角度+機械手臂 ➔ 提高AoA/AoD有效角度
- ✓ 雙圓極化+time gating ➔ 解決AoA=AoD
- ✓ Baffle ➔ 解決散射問題





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(臨廣科技產業園區)

